

# How to Write Fast Numerical Code

Spring 2013

*Lecture:* Discrete Fourier transform, fast Fourier transforms

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## Rest of Semester

May 2013

Sunday	Monday	Tuesday	Wednesday	Thursday	Friday	Saturday
			1	2	3	4
	6	7	8	9	10	11
	13	14	15	16	17	18
		21	22	23	24	25
	27	28	29	30	31	



Today



Lecture



Project meetings



Project presentations

- 10 minutes each
- random order
- random speaker

# Linear Transforms

- Overview: Transforms and algorithms
- Discrete Fourier transform
- Fast Fourier transforms
- After that:
  - Optimized implementation and autotuning (FFTW)
  - Automatic program synthesis (Spiral)

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# FFT References

- Complexity: Bürgisser, Clausen, Shokrollahi, *Algebraic Complexity Theory*, Springer, 1997
- History: Heideman, Johnson, Burrus: *Gauss and the History of the Fast Fourier Transform*, Arch. Hist. Sc. 34(3) 1985
- FFTs:
  - Cooley and Tukey, *An algorithm for the machine calculation of complex Fourier series*, "Math. of Computation, vol. 19, pp. 297–301, 1965
  - Nussbaumer, *Fast Fourier Transform and Convolution Algorithms*, 2nd ed., Springer, 1982
  - van Loan, *Computational Frameworks for the Fast Fourier Transform*, SIAM, 1992
  - Tolimieri, An, Lu, *Algorithms for Discrete Fourier Transforms and Convolution*, Springer, 2nd edition, 1997
  - Franchetti, Püschel, Voronenko, Chellappa and Moura, *Discrete Fourier Transform on Multicore*, IEEE Signal Processing Magazine, special issue on "Signal Processing on Platforms with Multiple Cores", Vol. 26, No. 6, pp. 90-102, 2009

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# Linear Transforms

- Very important class of functions: signal processing, scientific computing, ...
- **Mathematically:** Change of basis = Multiplication by a fixed matrix  $T$

$$\begin{pmatrix} y_0 \\ y_1 \\ \vdots \\ y_{n-1} \end{pmatrix} = y = Tx \quad T \cdot \quad x = \begin{pmatrix} x_0 \\ x_1 \\ \vdots \\ x_{n-1} \end{pmatrix}$$

$T = [t_{k,\ell}]_{0 \leq k, \ell < n}$

**Output**   **Input**

- Equivalent definition: Summation form

$$y_k = \sum_{\ell=0}^{n-1} t_{k,\ell} x_\ell, \quad 0 \leq k < n$$

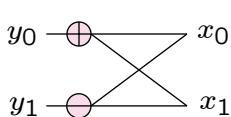
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## Smallest Relevant Example: DFT, Size 2

Transform (matrix):  $T = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$

Computation:  $y = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix} x$       or       $\begin{aligned} y_0 &= x_0 + x_1 \\ y_1 &= x_0 - x_1 \end{aligned}$

As graph (direct acyclic graph or DAG):



**called a butterfly**



[http://charlottesmartypons.blogspot.com/2011\\_02\\_01\\_archive.html](http://charlottesmartypons.blogspot.com/2011_02_01_archive.html)

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## Transforms: Examples

- A few dozen transforms are relevant
- Some examples

$$\text{DFT}_n = [e^{-2k\ell\pi i/n}]_{0 \leq k, \ell < n}$$

$$\text{RDFT}_n = [r_{k\ell}]_{0 \leq k, \ell < n}, \quad r_{k\ell} = \begin{cases} \cos \frac{2\pi k\ell}{n}, & k \leq \lfloor \frac{n}{2} \rfloor \\ -\sin \frac{2\pi k\ell}{n}, & k > \lfloor \frac{n}{2} \rfloor \end{cases}$$

*universal tool*

$$\text{DHT} = [\cos(2k\ell\pi/n) + \sin(2k\ell\pi/n)]_{0 \leq k, \ell < n}$$

$$\text{WHT}_n = \begin{bmatrix} \text{WHT}_{n/2} & \text{WHT}_{n/2} \\ \text{WHT}_{n/2} & -\text{WHT}_{n/2} \end{bmatrix}, \quad \text{WHT}_2 = \text{DFT}_2$$

$$\text{IMDCT}_n = [\cos((2k+1)(2\ell+1+n)\pi/4n)]_{0 \leq k < 2n, 0 \leq \ell < n}$$

$$\text{DCT-2}_n = [\cos(k(2\ell+1)\pi/2n)]_{0 \leq k, \ell < n}$$

*MPEG*

*JPEG*

$$\text{DCT-3}_n = \text{DCT-2}_n^T \quad (\text{transpose})$$

$$\text{DCT-4}_n = [\cos((2k+1)(2\ell+1)\pi/4n)]_{0 \leq k, \ell < n}$$

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## Blackboard

- Discrete Fourier transform (DFT)
- Transform algorithms
- Fast Fourier transform, size 4

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## Linear Transforms: DFT

$$\begin{pmatrix} y_0 \\ y_1 \\ \vdots \\ y_{n-1} \end{pmatrix} = y = Tx \quad T \cdot \quad x = \begin{pmatrix} x_0 \\ x_1 \\ \vdots \\ x_{n-1} \end{pmatrix}$$

*Output*   *Input*

**Example:**  $T = \text{DFT}_n = [e^{-2k\ell\pi i/n}]_{0 \leq k, \ell < n}$

$$= [\omega_n^{k\ell}]_{0 \leq k, \ell < n}, \quad \omega_n = e^{-2\pi i/n}$$

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## Algorithms: Example FFT, $n = 4$

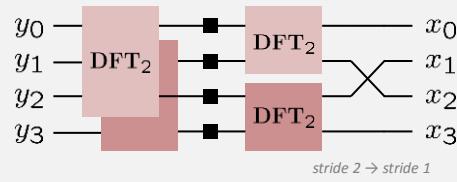
### Fast Fourier transform (FFT)

$$\begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & i & -1 & -i \\ 1 & -1 & 1 & -1 \\ 1 & -i & -1 & i \end{bmatrix} x = \begin{bmatrix} 1 & & 1 & & \\ & 1 & & 1 & \\ & & 1 & & \\ & & & 1 & \\ & & & & -1 \end{bmatrix} \begin{bmatrix} 1 & & & & \\ & 1 & & & \\ & & 1 & & \\ & & & 1 & \\ & & & & i \end{bmatrix} \begin{bmatrix} 1 & & 1 & & \\ & 1 & & -1 & \\ & & 1 & & \\ & & & 1 & \\ & & & & -1 \end{bmatrix} \begin{bmatrix} 1 & & & & \\ & 1 & & & \\ & & 1 & & \\ & & & 1 & \\ & & & & 1 \end{bmatrix} x$$

### Representation using matrix algebra

$$\text{DFT}_4 = (\text{DFT}_2 \otimes \text{I}_2) \text{ diag}(1, 1, 1, i) (\text{I}_2 \otimes \text{DFT}_2) L_2^4$$

### Data flow graph



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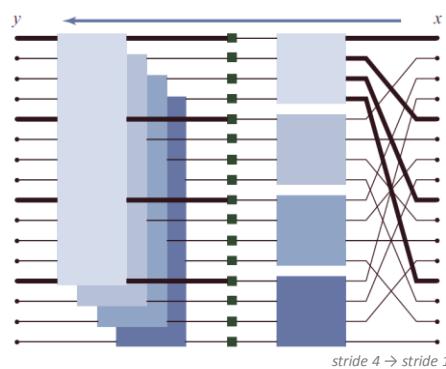
## Cooley-Tukey FFT (Recursive, General-Radix)

- Blackboard
- Kronecker products
- Stride permutations

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## Example FFT, n = 16 (Recursive, Radix 4)

$$\text{DFT}_{16} = \begin{matrix} & \text{DFT}_4 \otimes I_4 & T_4^{16} & I_4 \otimes \text{DFT}_4 & L_4^{16} \end{matrix}$$



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## Recursive Cooley-Tukey FFT

$$\text{DFT}_{km} = (\text{DFT}_k \quad I_m) T_m^{km} (I_k \quad \text{DFT}_m) L_k^{km} \quad \text{decimation-in-time}$$

*radix*

$$\text{DFT}_{km} = L_m^{km} (I_k \quad \text{DFT}_m) T_m^{km} (\text{DFT}_k \quad I_m) \quad \text{decimation-in-frequency}$$

- For powers of two  $n = 2^t$  sufficient together with base case

$$\text{DFT}_2 = \begin{bmatrix} 1 & 1 \\ 1 & -1 \end{bmatrix}$$

- Cost:

- (complex adds, complex mults) =  $(n \log_2(n), n \log_2(n)/2)$   
*independent of recursion*
- (real adds, real mults)  $\leq (2n \log_2(n), 3n \log_2(n)) = 5n \log_2(n)$  flops  
*depends on recursion: best is at least radix-8*

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## Recursive vs. Iterative FFT

- Recursive, radix-k Cooley-Tukey FFT

$$\text{DFT}_{km} = (\text{DFT}_k \quad I_m) T_m^{km} (I_k \quad \text{DFT}_m) L_k^{km}$$

$$\text{DFT}_{km} = L_m^{km} (I_k \quad \text{DFT}_m) T_m^{km} (\text{DFT}_k \quad I_m)$$

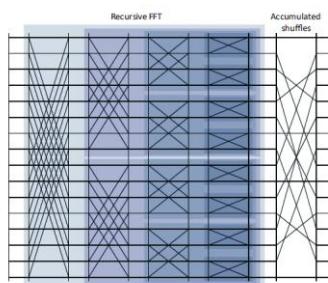
- Iterative, radix 2, decimation-in-time/decimation-in-frequency

$$\text{DFT}_{2^t} = \left( \prod_{j=1}^t (I_{2^{j-1}} \quad \text{DFT}_2 \quad I_{2^{t-j}}) \cdot (I_{2^{j-1}} \quad T_{2^{t-j}}^{2^{t-j+1}}) \right) \cdot R_{2^t}$$

$$\text{DFT}_{2^t} = R_{2^t} \cdot \left( \prod_{j=1}^t (I_{2^{t-j}} \quad T_{2^{j-1}}^{2^j}) \cdot (I_{2^{t-j}} \quad \text{DFT}_2 \quad I_{2^{j-1}}) \right)$$

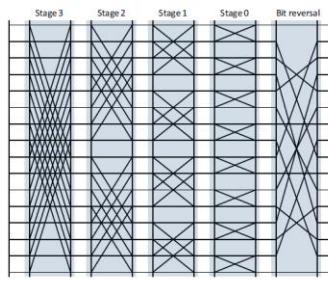
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### Radix 2, recursive



$$(\text{DFT}_2 \otimes I_8)T_8^{16} \left( I_2 \otimes \left( (\text{DFT}_2 \otimes I_4)T_4^8 (I_2 \otimes ((\text{DFT}_2 \otimes I_2)T_2^4 (I_2 \otimes \text{DFT}_2) L_2^4)) L_2^8 \right) \right) L_2^{16}$$

### Radix 2, iterative



$$\left( (I_1 \otimes \text{DFT}_2 \otimes I_8) D_0^{16} \right) \left( (I_2 \otimes \text{DFT}_2 \otimes I_4) D_1^{16} \right) \left( (I_4 \otimes \text{DFT}_2 \otimes I_2) D_2^{16} \right) \left( (I_8 \otimes \text{DFT}_2 \otimes I_1) D_3^{16} \right) R_2^{16}$$

## Recursive vs. Iterative

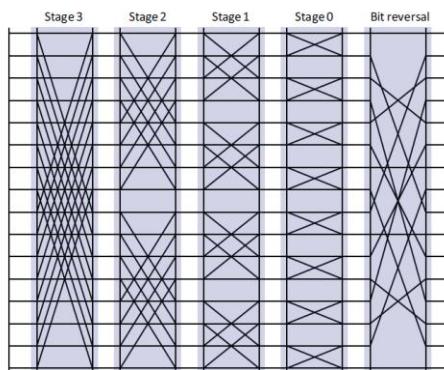
- **Iterative FFT computes in stages of butterflies =  $\log_2(n)$  passes through the data**
- **Recursive FFT reduces passes through data = better locality**
- **Same computation graph but different topological sorting**
- **Rough analogy:**

MMM	DFT
Triple loop	Iterative FFT
Blocked	Recursive FFT

## The FFT Is Very Malleable

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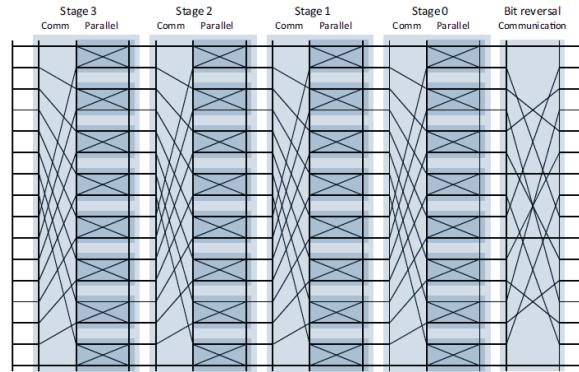
## Iterative FFT, Radix 2



$$\left( (I_1 \otimes \text{DFT}_2 \otimes I_8) D_0^{16} \right) \left( (I_2 \otimes \text{DFT}_2 \otimes I_4) D_1^{16} \right) \left( (I_4 \otimes \text{DFT}_2 \otimes I_2) D_2^{16} \right) \left( (I_8 \otimes \text{DFT}_2 \otimes I_1) D_3^{16} \right) R_2^{16}$$

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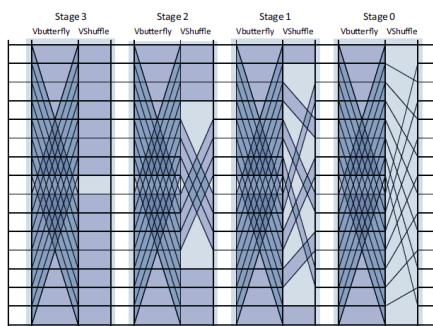
## Pease FFT, Radix 2



$$\left( L_2^{16} (I_8 \otimes DFT_2) D_0^{16} \right) \left( L_2^{16} (I_8 \otimes DFT_2) D_1^{16} \right) \left( L_2^{16} (I_8 \otimes DFT_2) D_2^{16} \right) \left( L_2^{16} (I_8 \otimes DFT_2) D_3^{16} \right) R_2^{16}$$

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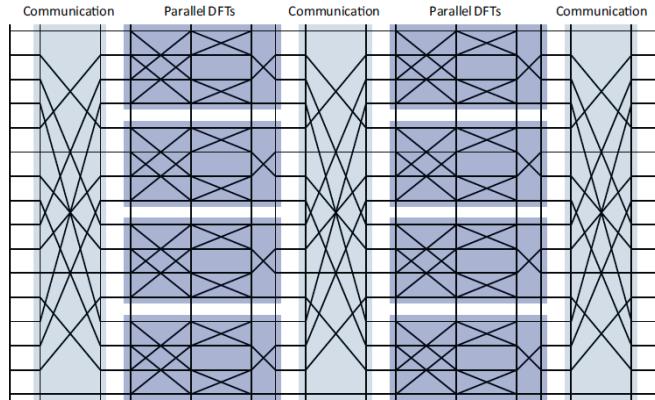
## Stockham FFT, Radix 2



$$\left( (DFT_2 \otimes I_8) D_0^{16} (L_2^2 \otimes I_8) \right) \left( (DFT_2 \otimes I_8) D_1^{16} (L_2^4 \otimes I_4) \right) \left( (DFT_2 \otimes I_8) D_2^{16} (L_2^8 \otimes I_2) \right) \left( (DFT_2 \otimes I_8) D_3^{16} (L_2^{16} \otimes I_1) \right)$$

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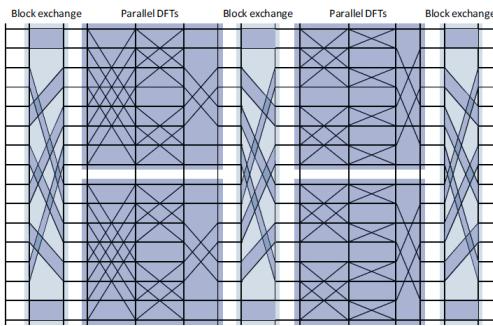
## Six-Step FFT



$$L_4^{16} \left( I_4 \otimes ((DFT_2 \otimes I_2) T_2^4 (I_2 \otimes DFT_2) L_2^4) \right) L_4^{16} T_4^{16} \left( I_4 \otimes ((DFT_2 \otimes I_2) T_2^4 (I_2 \otimes DFT_2) L_2^4) \right) L_4^{16}$$

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## Multi-Core FFT



$$(L_4^8 \otimes I_2) \left( I_2 \otimes ((DFT_2 \otimes I_2) T_2^4 (I_2 \otimes DFT_2) L_2^4) \otimes I_2 \right) (L_2^8 \otimes I_2) T_4^{16} \left( I_2 \otimes (I_2 \otimes (DFT_2 \otimes I_2) T_2^4 (I_2 \otimes DFT_2)) R_2^8 \right) (L_2^8 \otimes I_2)$$

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# Transform Algorithms

$$\begin{aligned}
& \text{DFT}_n \rightarrow P_{k/2,2m}^\top (\text{DFT}_{2m} \oplus (I_{k/2-1-i} C_{2m} r\text{DFT}_{2m}(i/k))) (r\text{DFT}'_k \quad I_m), \quad k \text{ even}, \\
& \begin{vmatrix} r\text{DFT}'_n \\ r\text{DFT}'_n \\ \text{DHT}_n \\ \text{DHT}'_n \end{vmatrix} \rightarrow (P_{k/2,m}^\top \quad I_2) \left( \begin{pmatrix} r\text{DFT}'_{2m} \\ r\text{DFT}'_{2m} \\ \text{DHT}'_{2m} \\ \text{DHT}'_{2m} \end{pmatrix} \oplus \begin{pmatrix} |r\text{DFT}_{2m}(i/k)| \\ |r\text{DFT}_{2m}(i/k)| \\ |r\text{DHT}_{2m}(i/k)| \\ |r\text{DHT}_{2m}(i/k)| \end{pmatrix} \right) \begin{pmatrix} r\text{DFT}'_k \\ r\text{DFT}'_k \\ \text{DHT}'_k \\ \text{DHT}'_k \end{pmatrix} \quad I_m, \quad k \text{ even}, \\
& |r\text{DFT}_{2n}(u)| \rightarrow L_{2n}^2 \left( I_k \quad i \begin{vmatrix} |r\text{DFT}_{2m}((i+u)/k)| \\ |r\text{DHT}_{2m}((i+u)/k)| \end{vmatrix} \right) \begin{pmatrix} |r\text{DFT}_{2k}(u)| \\ |r\text{DHT}_{2k}(u)| \end{pmatrix} \quad I_m, \\
& r\text{DFT}_{-3n} \rightarrow (Q_{k/2,m}^\top \quad I_2) (I_k \quad i r\text{DFT}_{2m}(i+1/2/k)) (r\text{DFT}-3_k \quad I_m), \quad k \text{ even}, \\
& \text{DCT-2}_n \rightarrow P_{k/2,2m}^\top (\text{DCT-2}_m K_2^{2m} \oplus (I_{k/2-1} \quad N_{2m} r\text{DFT}-3_{2m}^\top)) B_n (L_{k/2}^{n/2} \quad I_2) (I_m \quad r\text{DFT}'_k) Q_{m/2,k}, \\
& \text{DCT-3}_n \rightarrow \text{DCT-2}_n^\top, \\
& \text{DCT-4}_n \rightarrow Q_{k/2,2m}^\top (I_k/2 \quad N_{2m} r\text{DFT}-3_{2m}^\top) B'_n (L_{k/2}^{n/2} \quad I_2) (I_m \quad r\text{DFT}-3_k) Q_{m/2,k}. \\
& \text{DFT}_n \rightarrow (\text{DFT}_k \quad I_m) T_m^n (I_k \quad \text{DFT}_m) L_k^n, \quad n = km \xrightarrow{\text{Cooley-Tukey FFT}} \\
& \text{DFT}_n \rightarrow P_n (\text{DFT}_k \quad \text{DFT}_m) Q_n, \quad n = km, \quad \gcd(k, m) = 1 \xrightarrow{\text{Prime-factor FFT}} \\
& \text{DFT}_p \rightarrow R_p^T (I_1 \oplus \text{DFT}_{p-1}) D_p (I_1 \oplus \text{DFT}_{p-1}) R_p, \quad p \text{ prime} \xrightarrow{\text{Rader FFT}} \\
& \text{DCT-3}_n \rightarrow (I_m \oplus J_m) L_m^n (\text{DCT-3}_m(1/4) \oplus \text{DCT-3}_m(3/4)) \\
& \quad \cdot (F_2 \quad I_m) \begin{bmatrix} I_m & 0 \oplus -J_{m-1} \\ 0 \oplus J_{m-1} & \frac{1}{\sqrt{2}} (I_1 \oplus 2I_m) \end{bmatrix}, \quad n = 2m \\
& \text{DCT-4}_n \rightarrow S_n \text{DCT-2}_n \text{diag}_{0 \leq k < n} (1/(2 \cos((2k+1)\pi/4n))) \\
& \text{IMDCT}_{2m} \rightarrow (J_m \oplus I_m \oplus I_m \oplus J_m) \left( \begin{bmatrix} 1 \\ -1 \end{bmatrix} \quad I_m \right) \oplus \left( \begin{bmatrix} -1 \\ -1 \end{bmatrix} \quad I_m \right) J_{2m} \text{DCT-4}_{2m} \\
& \text{WHT}_{2^k} \rightarrow \prod_{i=1}^t (I_{2^{k_1+\dots+k_{i-1}}} \quad \text{WHT}_{2^{k_i}} \quad I_{2^{k_{i+1}+\dots+k_t}}), \quad k = k_1 + \dots + k_t \\
& \text{DFT}_2 \rightarrow F_2 \\
& \text{DCT-2}_2 \rightarrow \text{diag}(1, 1/\sqrt{2}) F_2 \\
& \text{DCT-4}_2 \rightarrow J_2 R_{13\pi/8}
\end{aligned}$$

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# Complexity of the DFT

- **Measure:**  $L_c$ ,  $2 \leq c$ 
  - Complex adds count 1
  - Complex mult by a constant  $a$  with  $|a| < c$  counts 1
  - $L_2$  is strictest,  $L_\infty$  the loosest (and most natural)
- **Upper bounds:**
  - $n = 2^k$ :  $L_2(\text{DFT}_n) \leq 3/2 n \log_2(n)$  *(using Cooley-Tukey FFT)*
  - General  $n$ :  $L_2(\text{DFT}_n) \leq 8 n \log_2(n)$  *(needs Bluestein FFT)*
- **Lower bound:**
  - Theorem by Morgenstern: If  $c < \infty$ , then  $L_c(\text{DFT}_n) \geq \frac{1}{2} n \log_c(n)$
  - Implies: in the measure  $L_c$ , the DFT is  $\Theta(n \log(n))$

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## History of FFTs

- The advent of digital signal processing is often attributed to the FFT  
(Cooley-Tukey 1965)
- History:
  - Around 1805: FFT discovered by Gauss [1]  
(Fourier publishes the concept of Fourier analysis in 1807!)
  - 1965: Rediscovered by Cooley-Tukey

[1]: Heideman, Johnson, Burrus: "Gauss and the History of the Fast Fourier Transform" Arch. Hist. Sc. 34(3) 1985<sup>25</sup>

## Carl-Friedrich Gauss



1777 - 1855

- Contender for the greatest mathematician of all times
- Some contributions: Modular arithmetic, least square analysis, normal distribution, fundamental theorem of algebra, Gauss elimination, Gauss quadrature, Gauss-Seidel, non-Euclidean geometry, ...

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