

Application to Linear Systems:

Given the system:

$$\left| \begin{array}{l} \dot{\underline{x}} = A\underline{x} + B\underline{u} \\ \underline{y} = C\underline{x} + D\underline{u} \end{array} \right| \quad (\text{MIMO})$$

with state- and input weighting:

$$PI = \underline{x}'(t_f) S \underline{x}(t_f) + \int_0^{t_f} \{ \underline{x}'(t) \cdot Q \underline{x}(t) + \underline{u}'(t) R \underline{u}(t) \} dt \stackrel{!}{=} \underset{\underline{u}(t)}{\text{Min}}$$

where: $S \geq \phi$; $Q \geq \phi$; $R > \phi$

Assume: $t_f = \underline{\text{fixed}}$; $\underline{x}(t_f) = \underline{\text{variable}}$;
 $\underline{u}(t) = \underline{\text{unlimited}}$

Q :: positive semidefinite (symmetric)
state-weighting matrix

R :: positive definite (symmetric)
input-weighting matrix

S :: positive semidefinite final-
value-weighting matrix

We can build the Hamiltonian:

$$H(\underline{x}, \underline{u}, \underline{\psi}, t) = \underline{x}' Q \underline{x} + \underline{u}' R \underline{u} + \underline{\psi}' (A \underline{x} + B \underline{u})$$

$$= \underline{x}' Q \underline{x} + \underline{\psi}' A \underline{x} + \underline{\psi}' B \underline{u} + \underline{u}' R \underline{u}$$

$$\Rightarrow \dot{\underline{\psi}} = - \frac{\partial H}{\partial \underline{x}} = -2Q \underline{x} - A' \underline{\psi}$$

where: $\underline{\psi}(t_f) = 2S \underline{x}(t_f)$

$$\frac{\partial H}{\partial \underline{u}} = 0 = 2R \underline{u}_{opt} + B' \underline{\psi}$$

$$\Rightarrow \underline{u}_{opt}(t) = -\frac{1}{2} R^{-1} B' \underline{\psi}$$

$$\dot{\underline{x}} = A \underline{x} + B \underline{u}_{opt} = A \underline{x} - \frac{1}{2} B R^{-1} B' \underline{\psi}$$

\Rightarrow We must solve the following boundary value problem:

$$\begin{bmatrix} \dot{\underline{x}} \\ \dot{\underline{\psi}} \end{bmatrix} = \begin{bmatrix} A & -\frac{1}{2} B R^{-1} B' \\ -2Q & -A' \end{bmatrix} \begin{bmatrix} \underline{x} \\ \underline{\psi} \end{bmatrix}$$

$$\left| \begin{array}{l} \underline{x}(0) = \underline{x}_0 \\ \underline{\psi}(t_f) = 2S \underline{x}(t_f) \end{array} \right|$$

We can simplify this a little bit by making:

$$\tilde{\psi}(t) = \frac{1}{2} \cdot \psi(t)$$

$$\Leftrightarrow \psi(t) = 2 \cdot \tilde{\psi}(t)$$

$$\Rightarrow \dot{\psi}(t) = 2 \cdot \dot{\tilde{\psi}}(t)$$

$$\Rightarrow \begin{bmatrix} \dot{x} \\ \dot{\tilde{\psi}} \end{bmatrix} = \begin{bmatrix} A & -BR^{-1}B' \\ -Q & -A' \end{bmatrix} \cdot \begin{bmatrix} x \\ \tilde{\psi} \end{bmatrix}$$

$$\left| \begin{array}{l} x(0) = x_0 \\ \tilde{\psi}(t_f) = Sx(t_f) \end{array} \right|$$

This $(2n \times 2n)$ "system" matrix is called the Hamiltonian matrix.

Solution:

We write for $\tilde{\psi}(t)$:

$$\boxed{\tilde{\psi}(t) = P(t) \cdot x(t)}$$

As nothing was said about

$P(t)$, this is obviously possible.

$$\Rightarrow \underline{\psi}(t_f) = P(t_f) \cdot \underline{x}(t_f) \equiv S \underline{x}(t_f)$$

$$\Rightarrow \boxed{P(t_f) \equiv S}$$

$$\underline{\dot{x}} = A \underline{x} - BR^{-1}B' \underline{\psi}$$

$$= A \underline{x} - BR^{-1}B' P \underline{x}$$

$$\Rightarrow \underline{\dot{x}} = [A - BR^{-1}B'P] \underline{x}$$

$$\underline{\dot{\psi}} = -Q \underline{x} - A' \underline{\psi}$$

$$= -Q \underline{x} - A' P \underline{x}$$

$$\Rightarrow \underline{\dot{\psi}} = -[Q + A'P] \underline{x}$$

but: $\underline{\psi} = P \underline{x}$

$$\Rightarrow \underline{\dot{\psi}} = (P \underline{x}) \dot{=} \dot{P} \underline{x} + P \underline{\dot{x}}$$

$$-[Q + A'P] \underline{x} = \dot{P} \underline{x} + P[A - BR^{-1}B'P] \underline{x}$$

$$\Rightarrow \boxed{-\dot{P} = PA + A'P + Q - PBR^{-1}B'P}$$

This is called the Riccati
Differential Equation.

⇒ We have decomposed the
(2n) - boundary value problem
into two initial value
problems, one of size (n), the
other of size (n × n).

Recipe: Integrate the Riccati
equation backward in time
from $t_f \rightarrow 0$:

$$\dot{P} = PBR^{-1}B'P - PA - A'P - Q$$

$$P(t_f) = S$$

⇒ P(t)

Then plug the previously found
P(t) into the system equations
and integrate them forward
in time from $0 \rightarrow t_f$:

$$\left| \begin{aligned} \dot{\underline{x}} &= (A - BR^{-1}B'P(t)) \underline{x} \\ \underline{x}(\varphi) &= \underline{x}_0 \end{aligned} \right|$$

$$\Rightarrow \underline{x}(t)$$

Notice:

$$\begin{aligned} \underline{u}_{opt}(t) &= -\frac{1}{2} R^{-1} B' \underline{\psi}(t) \\ &= -R^{-1} B' \tilde{\underline{\psi}}(t) \end{aligned}$$

$$\Rightarrow \underline{u}_{opt}(t) = -\underbrace{R^{-1} B' P(t)}_{K(t)} \cdot \underline{x}(t)$$

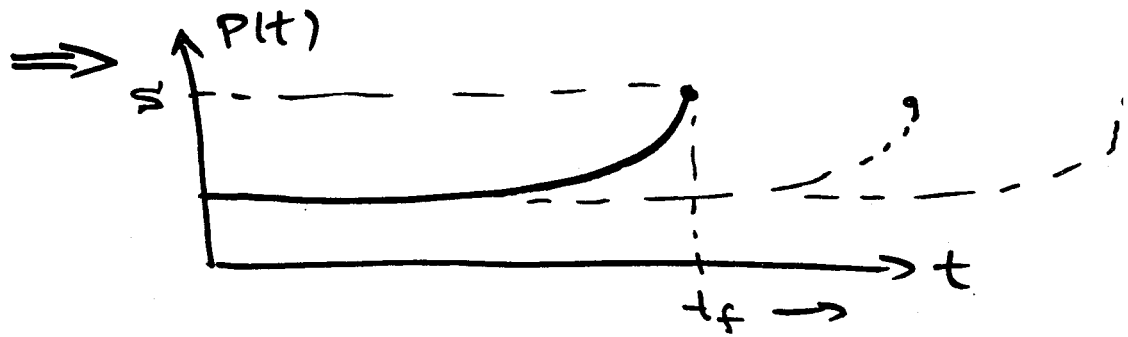
$$\Rightarrow \underline{u}_{opt}(t) = \underline{r}(t) - K(t) \cdot \underline{x}(t)$$

where: $K(t) = R^{-1} B' P(t)$

The optimal solution can be realized as a time-variant state-feedback.

Simplification:

$$t_f \rightarrow \infty$$



$\Rightarrow P(t)$ becomes constant.

$$\Rightarrow \dot{P} = 0$$

$$\Rightarrow PA + A'P + Q - PBR^{-1}B'P = 0$$

\Rightarrow Algebraic Matrix-Riccati Equation.

$$\Rightarrow K = R^{-1}B'P$$

$$\Rightarrow \underline{u}_{opt}(t) = \underline{r}(t) - K \cdot \underline{x}(t)$$

is a state-feedback.

- The "algebraic Matrix-Riccati Equation" is actually a set of n^2 nonlinear equations.

- This set of equations has many solutions, but only one leads to a stable feedback system.

Algorithm: (without proof)

① Check controllability. If not completely controllable, input-decouple the uncontrollable modes first.

② Compute the Hamiltonian matrix:

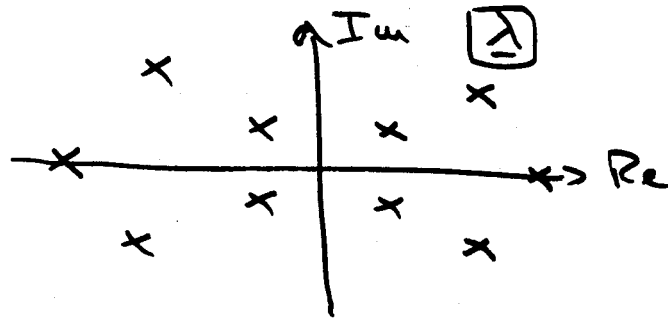
$$H = \begin{bmatrix} A & -BR^{-1}B' \\ -Q & -A' \end{bmatrix}$$

③ Compute the spectral decomposition of H :

$$[V, \lambda] = \text{Eig}(H)$$

The eigenvalues of the Hamiltonian are not only symmetric to the real axis, but also to the imaginary axis.

If the system is controllable
 $\Rightarrow H$ has no eigenvalues on
 the imaginary axis. E.g.:



might be a possible set of
 eigenvalues of the Hamiltonian.

\Rightarrow There are exactly n eigenvalues
 with negative real part.

- ④ Take the subset of the right
 Modal matrix that relates to
 eigenvalues with negative real
 part:

$$\hat{V} = \begin{matrix} n \\ \boxed{} \\ 2n \end{matrix}$$

- ⑤ Cut \hat{V} into an upper and
 a lower portion:

$$\hat{V} = \begin{bmatrix} V_1 \\ V_2 \end{bmatrix} = \begin{array}{|c} \hline V_1 \\ \hline V_2 \\ \hline \end{array}$$

⑥ Compute the solution to the algebraic Riccati equation:

$$P = V_2 V_1^{-1}$$

(Notice: Any other n eigenvectors would have given us another solution, but this is the one and only solution that leads to a stable closed-loop system.)

⑦ $K = R^{-1} B' P$

is the desired state feedback.

This algorithm is numerically very benign (much better than pole placement).

In Matlab:

```
H = [A, -(B/R)*B'; -Q, -A'];  
[V, L] = eig(H);  
k = 0;  
for i = 1:2*n,  
    if real(L(i,i)) < 0 then  
        k = k + 1;  
        V(:,k) = V(:,i);  
    end  
end  
V1 = V(1:n, 1:n);  
V2 = V(n+1:2*n, 1:n);  
P = V2 / V1;  
K = (R \ (B')) * P;
```

This algorithm also exists as a preprogrammed Matlab function:

$$K = \text{lqr}(A, B, Q, R)$$

↳ linear quadratic (Gaussian) regulator

Disadvantages: LQG has a tendency of placing poles too close to each other \Rightarrow large sensitivity to parameter changes.

Solution: There meanwhile exist techniques to individually influence pole locations in the Riccati design \Rightarrow mixture between Riccati & pole placement \Rightarrow probably better than any of the two alone \Rightarrow Hal Tharp

- LQG and PLACE can both be used for state-feedback design. \Rightarrow We can obtain output feedback e.g. by solving two LQG-problems one for the controller, and

one for the observer:

$$\Rightarrow P = LQR(A, B, Q_c, R_c);$$

$$\Rightarrow P = LQR(A', C', Q_o, R_o);$$

$$\Rightarrow R = R';$$

Output Weighting:

Sometimes, it is desirable to weight the outputs instead of the states:

$$\left| \begin{array}{l} \dot{\underline{x}} = A\underline{x} + B\underline{u} \\ \underline{y} = C\underline{x} \end{array} \right| \quad (\underline{D} = \underline{\emptyset})$$

$$PI = \int_0^{\infty} \{ \underline{y}' Q \underline{y} + \underline{u}' R \underline{u} \} dt \stackrel{!}{=} \text{Min}_{\underline{u}(t)}$$

$$Q \geq \emptyset ; R > \emptyset$$

$$\Rightarrow \underline{u}' Q \underline{y} = (C\underline{x})' Q (C\underline{x})$$

$$= \underline{x}' \underbrace{C' Q C}_{Q_n} \underline{x} ; Q_n \geq \emptyset$$

$$\Rightarrow PI = \int_0^{\infty} \{ \underline{x}' Q_n \underline{x} + \underline{u}' R \underline{u} \} dt \stackrel{!}{=} \text{Min}_{\underline{u}(t)}$$

is an equivalent problem with state weighting.

• The case $D \neq \emptyset$ works also, but is a little more tricky.

$$\begin{aligned} \underline{y}' Q \underline{y} &= (\underline{C} \underline{x} + \underline{D} \underline{u})' Q (\underline{C} \underline{x} + \underline{D} \underline{u}) \\ &= (\underline{x}' \underline{C}' + \underline{u}' \underline{D}') Q (\underline{C} \underline{x} + \underline{D} \underline{u}) \\ &= \underline{x}' \underline{C}' Q \underline{C} \underline{x} + \underline{x}' \underline{C}' Q \underline{D} \underline{u} + \underline{u}' \underline{D}' Q \underline{C} \underline{x} \\ &\quad + \underline{u}' \underline{D}' Q \underline{D} \underline{u} \end{aligned}$$

Let:
$$\left| \begin{array}{l} \hat{Q} = \underline{C}' Q \underline{C} \\ \hat{R} = \underline{R} + \underline{D}' Q \underline{D} \\ \hat{N} = \underline{C}' Q \underline{D} \end{array} \right|$$

$$\Rightarrow PI = \int_0^{\infty} \{ \underline{x}' \hat{Q} \underline{x} + \underline{x}' \hat{N} \underline{u} + \underline{u}' \hat{N}' \underline{x} + \underline{u}' \hat{R} \underline{u} \}$$

$$\text{or: } PI = \int_0^{\infty} \begin{bmatrix} \underline{x}' & \underline{u}' \end{bmatrix} \cdot \begin{bmatrix} \hat{Q} & \hat{N} \\ \hat{N}' & \hat{R} \end{bmatrix} \cdot \begin{bmatrix} \underline{x} \\ \underline{u} \end{bmatrix} dt \stackrel{!}{=} \text{Min}_{\underline{u}(t)}$$

Another variable transformation can remove the mixed terms:

$$PI = \int_0^{\infty} \left\{ \underline{x}' (\hat{Q} - \hat{N} \hat{R}^{-1} \hat{N}') \underline{x} + (\underline{u} + \hat{R}^{-1} \hat{N}' \underline{x})' \cdot \hat{R} (\underline{u} + \hat{R}^{-1} \hat{N}' \underline{x}) \right\} dt \stackrel{!}{=} \underset{\underline{u}}{\text{Min}}$$

(Can be verified easily by multiplying out.)

$$\text{Let: } \left| \begin{array}{l} \underline{Q}_n = \hat{Q} - \hat{N} \hat{R}^{-1} \hat{N}' \\ \underline{u}_n = \underline{u} + \hat{R}^{-1} \hat{N}' \underline{x} \end{array} \right|$$

$$\Rightarrow PI = \int_0^{\infty} \left\{ \underline{x}' \underline{Q}_n \underline{x} + \underline{u}_n' \hat{R} \underline{u}_n \right\} dt \stackrel{!}{=} \underset{\underline{u}_n}{\text{Min}}$$

\Rightarrow is a performance index with state weighting and without mixed terms for a modified problem:

$$\left| \begin{array}{l} \dot{\underline{x}} = \underline{A}_n \underline{x} + \underline{B}_n \underline{u}_n \\ \underline{y} = \underline{C}_n \underline{x} + \underline{D}_n \underline{u}_n \end{array} \right|$$

$$\begin{aligned} \underline{y} &= C\underline{x} + D\underline{u} \\ &= C\underline{x} + D\underline{u}_n - D\hat{R}^{-1}\hat{N}'\underline{x} \\ \Rightarrow \underline{y} &= \underbrace{[C - D\hat{R}^{-1}\hat{N}']}_{C_n} \underline{x} + \underbrace{D}_{D_n} \underline{u}_n \end{aligned}$$

$$\begin{aligned} \dot{\underline{x}} &= A\underline{x} + B\underline{u} \\ &= A\underline{x} + B\underline{u}_n - B\hat{R}^{-1}\hat{N}'\underline{x} \\ \Rightarrow \dot{\underline{x}} &= \underbrace{[A - B\hat{R}^{-1}\hat{N}']}_{A_n} \underline{x} + \underbrace{B}_{B_n} \underline{u}_n \end{aligned}$$

$$\Rightarrow \left| \begin{array}{l} \dot{\underline{x}} = A_n \underline{x} + B \underline{u}_n \\ \underline{y} = C_n \underline{x} + D \underline{u}_n \end{array} \right|$$

where: $\left| \begin{array}{l} A_n = A - B\hat{R}^{-1}\hat{N}' \\ C_n = C - D\hat{R}^{-1}\hat{N}' \end{array} \right|$

$$PI = \int_0^{\infty} \{ \underline{x}' Q_n \underline{x} + \underline{u}_n' \hat{R} \underline{u}_n \} dt \stackrel{!}{=} \text{Min}_{\underline{u}_n}$$

with:

$$\left\{ \begin{array}{l} \hat{Q} = C'QC \\ \hat{N} = C'QD \\ \hat{R} = R + D'DQD \\ Q_s = \hat{Q} - \hat{N}\hat{R}^{-1}\hat{N}' \end{array} \right.$$

$$\Rightarrow \hat{H} = \begin{bmatrix} A_s & -B\hat{R}^{-1}B' \\ -Q_s & -A_s' \end{bmatrix}$$

$$\Rightarrow \dots \Rightarrow \hat{\pi} = B'\hat{R}^{-1}\hat{p}$$

$\hat{\pi}$ for new formulation

$$\Rightarrow u_s = r - \hat{\pi}'x \equiv u + \hat{R}^{-1}\hat{N}'x$$

$$\Rightarrow u = r - \underbrace{\left[\hat{\pi}' + \hat{R}^{-1}\hat{N}' \right]}_{\pi} x$$

$$\boxed{u = r - Kx}$$

is again a state feedback

where:

$$\boxed{K = \hat{\pi}' + \hat{R}^{-1}\hat{N}'}$$

$\hat{\pi}$ for old formulation

In Matlab:

Let us solve the problem:

$$\begin{cases} \dot{\underline{x}} = A\underline{x} + B\underline{u} \\ \underline{y} = C\underline{x} + D\underline{u} \end{cases}$$

$$PI = \int_0^{\infty} [\underline{x}' \quad \underline{u}'] \cdot \begin{bmatrix} \hat{Q} & \hat{N} \\ \hat{N}' & \hat{R} \end{bmatrix} \cdot \begin{bmatrix} \underline{x} \\ \underline{u} \end{bmatrix} dt = \min_{\underline{u}(t)}$$

Algorithm:

$$\begin{cases} A_n = A - (B/\hat{R}) * \hat{N}' \\ Q_n = \hat{Q} - (\hat{N}/\hat{R}) * \hat{N}' \\ \hat{K} = \text{lqr}(A_n, B, Q_n, \hat{R}); \\ K = \hat{K} + (\hat{R} \setminus \hat{N}'); \end{cases}$$

This also exists as a pre-programmed Matlab function:

$$K = \text{lqr}(A, B, \hat{Q}, \hat{R}, \hat{N});$$

Let us now solve the complete output weighting problem:

$$\begin{cases} \dot{\underline{x}} = A \cdot \underline{x} + B \cdot \underline{u} \\ \underline{y} = C \cdot \underline{x} + D \cdot \underline{u} \end{cases}$$

$$PI = \int_0^{\infty} [\underline{y}' \quad \underline{u}'] \cdot \begin{bmatrix} Q & \emptyset \\ \emptyset & R \end{bmatrix} \cdot \begin{bmatrix} \underline{y} \\ \underline{u} \end{bmatrix} dt \stackrel{!}{=} \min_{\underline{u}(t)}$$

Algorithm:

$$\begin{cases} \hat{Q} = C' * Q * C; \\ \hat{R} = R + D' * Q * D; \\ \hat{N} = C' * Q * D; \\ K = \text{lqr}(A, B, \hat{Q}, \hat{R}, \hat{N}); \end{cases}$$

This can also be obtained by:

$$\begin{cases} N = \text{zeros}(p, m); \\ K = \text{lqry}(A, B, C, D, Q, R, N); \end{cases}$$

↓ # of outputs
↓ # of inputs

Exponential Stability:

In order to guarantee a maximum settling time, we want all eigenvalues by at least a factor of σ left from the imaginary axis, where:

$$\sigma \approx \frac{4}{T_s}$$

With optimal control, this problem can also be solved very easily.

$$\left| \begin{array}{l} \dot{\underline{x}} = A\underline{x} + B\underline{u} \\ \underline{y} = C\underline{x} + D\underline{u} \end{array} \right|$$

be a system to be controlled.

• We can look at a different system:

$$\left| \begin{array}{l} \dot{\underline{x}}_n = \tilde{A}\underline{x}_n + B\underline{u} \\ \underline{y}_n = C\underline{x}_n + D\underline{u} \end{array} \right|$$

where: $\tilde{A} = A + \alpha I$

$$\Rightarrow \{ \text{Eig}(\tilde{A}) \} = \{ \text{Eig}(A) + \alpha \}$$

We now design the state feedback for the new system: K

$$\Rightarrow \tilde{A}_{cl} = \tilde{A} - BK$$

is stable.

• We now apply this feedback K to our original problem:

$$\Rightarrow A_{cl} = A - BK$$

$$\begin{aligned} \Rightarrow A_{cl} &= \tilde{A} - \alpha I - BK \\ &= \tilde{A} - BK - \alpha I \\ &= \tilde{A}_{cl} - \alpha I \end{aligned}$$

$$\Rightarrow \{ \text{Eig}(A_{cl}) \} = \{ \text{Eig}(\tilde{A}_{cl}) - \alpha \}$$

↑
at least α
left from
imaginary
axis.

↑
stable

Example:

Given the System:

$$\left| \begin{array}{l} \dot{\underline{x}} = \begin{bmatrix} -150 & 192 & 12 & 165 \\ 143 & -181 & -15 & -154 \\ -142 & 179 & 15 & 153 \\ -291 & 370 & 28 & 316 \end{bmatrix} \underline{x} + \begin{bmatrix} 1 \\ -1 \\ 1 \\ 2 \end{bmatrix} u \\ y = \begin{bmatrix} -27 & 51 & -18 & 48 \end{bmatrix} \underline{x} \end{array} \right|$$

(cf. page 232)

- We want to design an output feedback using optimal control where all controller poles are at least 4 from the imaginary axis, all observer poles are at least 8 from the imaginary axis. We use output weighting with:

$$Q = [1] \quad \text{and} \quad R = [1].$$

- We already have found a model for the system:

$$\left| \begin{array}{l} \underline{y} = \begin{bmatrix} \emptyset & 1 & \emptyset \\ \emptyset & \emptyset & 1 \\ -24 & 2 & 5 \end{bmatrix} \underline{y} + \begin{bmatrix} \emptyset \\ \emptyset \\ 1 \end{bmatrix} u \\ \underline{y} = \begin{bmatrix} 3 & 3 & \emptyset \end{bmatrix} \underline{y} \end{array} \right|$$

(cf. page 235)

- ⇒ [> A = [$\emptyset, 1, \emptyset$; $\emptyset, \emptyset, 1$; $-24, 2, 5$] ;
- [> B = [\emptyset ; \emptyset ; 1] ;
- [> C = [$3, 3, \emptyset$] ;
- [> D = \emptyset ;
- [> AT = A + 4 * EYE(A) ;
- [> Q = 1 ;
- [> R = 1 ;
- [> N = \emptyset ;
- [> K = LQR(A, B, C, D, Q, R, N)

$$\Rightarrow K = \begin{bmatrix} 769. \emptyset 225 & 272.1969 & 34. \emptyset \emptyset 79 \end{bmatrix}$$

$$[> \text{EIG}(A - B * K)$$

$$\Rightarrow \text{ANS} = \begin{bmatrix} -6. \emptyset \emptyset 42 \\ -11. \emptyset 395 \\ -11.9642 \end{bmatrix}$$

all real !!!

Observer design:

$$[> AT = A + 8 * \text{EYE}(A);$$

$$[> H = \text{LQR}(AT', C', B', D', Q, R, N);$$

$$[> H = H'$$

$$\Rightarrow H = \begin{bmatrix} -73.7788 \\ 93.1127 \\ 312.9170 \end{bmatrix}$$

$$[> \text{EIG}(A - H * C)$$

$$\rightarrow \text{ANS} = \begin{bmatrix} -14.0011 \\ -19.0153 \\ -19.9854 \end{bmatrix}$$

\Rightarrow \underline{k}' and \underline{h} are too large.

Balancing:

$$[> T = \text{SQRT}(\text{ABS}(K'./H));$$

$$[> T = \text{DIAG}(T);$$

$$[> AN = T * A / T;$$

$$[> BN = T * B;$$

$$[> CN = C / T;$$

$$\Rightarrow \begin{cases} \dot{\underline{y}} = \begin{bmatrix} \emptyset & 1.9893 & \emptyset \\ \emptyset & \emptyset & 5.1864 \\ -2.4507 & 0.3256 & 5 \end{bmatrix} \underline{y} + \begin{bmatrix} \emptyset \\ \emptyset \\ 0.3297 \end{bmatrix} u \\ \underline{y} = [0.9292 \quad 1.7546 \quad \emptyset] \underline{y} \end{cases}$$

is a new model.

$$\Rightarrow [> KN = K/T$$

$$\Rightarrow KN = \begin{bmatrix} 238.1965 & 159.2411 & 143.1583 \end{bmatrix}$$

$$[> HN = T^{-1}H$$

$$\Rightarrow HN = \begin{bmatrix} -238.1965 \\ 159.2411 \\ 143.1583 \end{bmatrix}$$

\Rightarrow We have requested too much.

Looking at the closed loop eigenvalues, we realize that LQR has placed them too far to the left.

- We repeat the design by lowering our demands:

controller poles left from -2.5
observer poles left from -5

(in the hope that LQR still places them where we want them to be).

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$$\Rightarrow \begin{aligned} & [> AT = A + 2.5 * EYE(A); \\ & [> K = LQRY(AT, B, C, D, Q, R, N) \end{aligned}$$

$$\Rightarrow K = [193.4091 \quad 125.3972 \quad 25.0227]$$

$$[> EIG(A - B * K)$$

$$\Rightarrow \text{ANS} = \begin{bmatrix} -3.014 \\ -8.0676 \\ -8.9411 \end{bmatrix} \quad \underline{\underline{\text{NO!}}}$$

\Rightarrow Iterate a little ...

$$[> AT = A + 3 * EYE(A);$$

$$[> K = LQRY(AT, B, C, D, Q, R, N)$$

$$\Rightarrow K = [337.1588 \quad 162.2872 \quad 28.0144]$$

$$[> EIG(A - B * K)$$

$$\Rightarrow \text{ANS} = \begin{bmatrix} -4.000 \\ -9.0554 \\ -9.9509 \end{bmatrix} \leftarrow \underline{\underline{\text{YES!}}}$$

$$[> AT = A + 6 * EYE(A);$$

$$[> H = LQRY(AT', C', B', D', Q, R, N);$$

$$[> H = H'$$

$$\Rightarrow H = \begin{bmatrix} -31.1798 \\ 46.5142 \\ 214.1975 \end{bmatrix}$$

$$[> EIG(A - H * C)$$

$$\rightarrow \text{ANS} = \begin{bmatrix} -10.0019 \\ -15.0232 \\ -15.9784 \end{bmatrix} \leftarrow \text{iterate a little ...}$$

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$$[> AT = A + S * EYE(A);$$

$$[> H = LQR4(AT', C', B', D', Q, R, N);$$

$$[> H = H'$$

$$\Rightarrow H = \begin{bmatrix} -17.8802 \\ 31.2152 \\ 168.8428 \end{bmatrix}$$

$$[> EIG(A - H * C)$$

$$\Rightarrow \text{ANS} = \begin{bmatrix} -8.0026 \\ -13.0297 \\ -13.9726 \end{bmatrix} \leftarrow \underline{\underline{\text{very good!}}}$$

Balancing:

$$[> T = \text{SQRT}(\text{ABS}(K' / H));$$

$$[> T = \text{DIAG}(T);$$

$$[> KN = K / T, \quad HN = T * H$$

$$\Rightarrow KN = [77.6431 \quad 72.4784 \quad 68.7752]$$

$$HN = \begin{bmatrix} -77.6431 \\ 72.4784 \\ 68.7752 \end{bmatrix}$$

This design looks okay.

The new model is:

$$\left| \begin{array}{l} \underline{\dot{y}} = \begin{bmatrix} 0 & 1.8702 & 0 \\ 0 & 0 & 5.7002 \\ -2.2513 & 0.3509 & 5 \end{bmatrix} \underline{y} + \begin{bmatrix} 0 \\ 0 \\ 0.4073 \end{bmatrix} \\ \underline{y} = [0.6909 \quad 1.292 \quad 0] \underline{u} \end{array} \right.$$

Verification:

$$\{> AN = T * A / T;$$

$$\{> BN = T * B;$$

$$\{> CN = C / T;$$

$$\{> AT = AN + S * EYE (AN);$$

$$\{> K = LQRY(AT, BN, CN, D, Q, R, N)$$

$$\Rightarrow K = \begin{bmatrix} 77.6431 & 72.4784 & 68.7752 \end{bmatrix}$$

✓

$$\{> AT = AN + S * EYE (AN);$$

$$\{> H = LQRY(AT', CN', BN', D', Q, R, N);$$

$$\{> H = H'$$

$$\Rightarrow H = \begin{bmatrix} -77.6431 \\ 72.4784 \\ 68.7752 \end{bmatrix}$$

✓

From this, we can learn that obviously optimality and displacement of eigenvalues (exponential stability) are invariant to this kind of similarity transformation.