

Problem: In the past, we had seen that two systems which differ in their transfer functions are different systems. Thus transfer functions are unique. Let us prove that two state-space representations which are similar to each other result in the same transfer function.

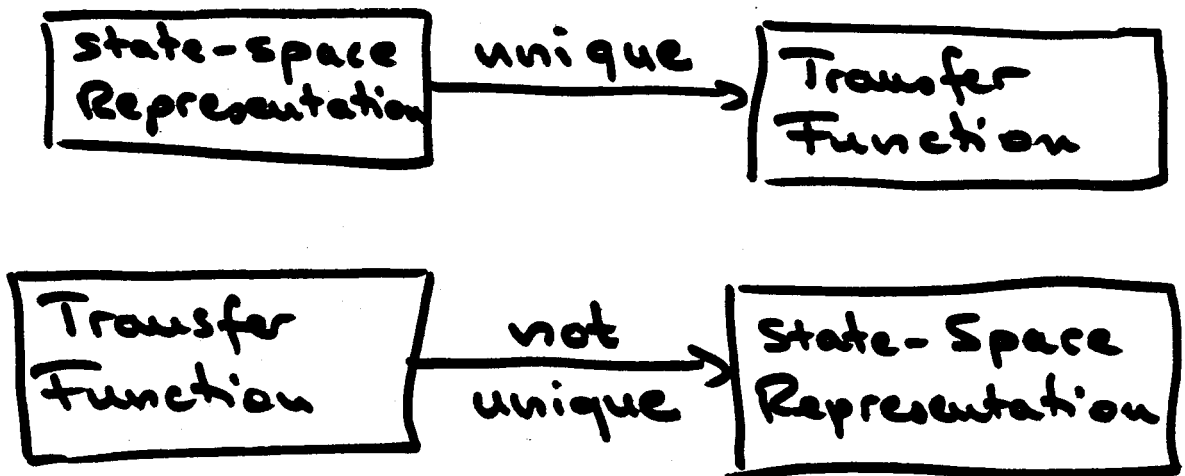
$$\left| \begin{array}{l} \dot{\underline{x}} = \underline{A}\underline{x} + \underline{b}u \\ y = \underline{c}'\underline{x} + du \end{array} \right| \Leftrightarrow G(s) = \underline{c}'(s\mathbf{I} - \underline{A})^{-1}\underline{b} + d$$

↓ similar

$$\left| \begin{array}{l} \dot{\underline{s}} = \underline{T}\underline{A}\underline{T}^{-1}\underline{s} + \underline{T}\underline{b}u \\ y = \underline{c}'\underline{T}^{-1}\underline{s} + du \end{array} \right| \Leftrightarrow \hat{G}(s) = ?$$

$$\begin{aligned}\hat{G}(s) &= \underline{\hat{C}}'(sI - \hat{A})^{-1}\hat{b} + \hat{d} \\ &\equiv \underline{C}'T^{-1}(sI - TAT^{-1})^{-1}Tb + d \\ &\equiv \underline{C}'T^{-1}(TsIT^{-1} - TAT^{-1})^{-1}Tb + d \\ &\equiv \underline{C}'T^{-1}\left[T(sI - A)T^{-1}\right]^{-1}Tb + d \\ &\equiv \underline{C}'T^{-1}T(sI - A)^{-1}T^{-1}Tb + d \\ &\equiv \underline{C}'(sI - A)^{-1}b + d \\ &\equiv G(s) \quad \text{q.e.d.}\end{aligned}$$

Thus :



Example :

$$\left. \begin{aligned} \dot{\underline{x}} &= \begin{bmatrix} -56 & -15 & 30 \\ 30 & 9 & -16 \\ -90 & -25 & 48 \end{bmatrix} \underline{x} + \begin{bmatrix} -2 \\ 1 \\ -3 \end{bmatrix} u \\ y &= \begin{bmatrix} -7 & -1 & 4 \end{bmatrix} \underline{x} \end{aligned} \right|$$

Find the transfer function !

$$Q_c = \left[ \underline{b}, A\underline{b}, A^2\underline{b} \right]$$

$$= \begin{bmatrix} -2 & 7 & -17 \\ 1 & -3 & 7 \\ -3 & 11 & -27 \end{bmatrix}$$

$$\Rightarrow \det(Q_c) \equiv 0$$

$\Rightarrow Q_c^{-1}$  does not exist !!!

Something went wrong. Let us try the other technique.

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$$(sI - A) = \begin{bmatrix} (s+56) & 15 & -30 \\ -30 & (s-9) & 16 \\ 90 & 25 & (s-48) \end{bmatrix}$$

$$\Rightarrow \det(sI - A) = s^3 - s^2 - 10s - 8$$

$$\text{adj}(sI - A) = \begin{bmatrix} (s^2 - 57s + 32) & (-15s - 30) & (30s - 16) \\ 30s & (s^2 + 8s + 12) & (-16s + 90) \\ (-90s + 60) & (-25s - 50) & (s^2 + 47s - 432) \end{bmatrix}$$

$$\Rightarrow (sI - A)^T \underline{b} = \begin{bmatrix} (-2s^2 + 9s - 4) \\ (s^2 - 4s) \\ (-3s^2 + 14s - 8) \end{bmatrix}$$

$$\Rightarrow \underline{c}' (sI - A)^T \underline{b} = (s^2 - 3s - 4)$$

$$\Rightarrow G(s) = \frac{s^2 - 3s - 4}{s^3 - s^2 - 10s - 8}$$

We factorize :

$$\underline{G(s)} = \frac{(s-4)(s+1)}{(s-4)(s+1)(s+2)} = \underline{\underline{\underline{s}}}$$

$\Rightarrow$  Pole/zero cancellations ,

$$\underline{\underline{G(s) = \frac{1}{s+2}}}$$

Thus, another state-space representation would be:

$$\left| \begin{array}{l} \dot{\xi} = -2\xi + u \\ y = \xi \end{array} \right|$$

Obviously,  $\xi$  is not similar to  $\underline{x}$  (even not dimensionwise)

- This system is obviously with respect to input/output behavior a first-order system. However, there exist some internal states that can either not be reached from the input (uncontrollable state) or that cannot be seen from the output (unobservable states)

Recipe: We must find a transformation which makes the uncontrollable states as such visible.

Algorithm:

- (1) Take as many columns of  $Q_c$  as are linearly independent, and extend by anything that gives a full rank:

$$\tilde{Q}_c = \left[ \begin{array}{c} Q_{c \text{ l.ind.}} \\ \vdots \\ \text{extension} \end{array} \right]$$

- (2) Use:

$$T = \tilde{Q}_c^{-1}$$

for a similarity transformation

Example continued:

$$Q_c = \begin{bmatrix} -2 & 7 & -17 \\ 1 & -3 & 7 \\ -3 & 11 & -27 \end{bmatrix}$$

↑ ↑  
lin. indep.

$$\Rightarrow \tilde{Q}_c = \begin{bmatrix} -2 & 7 & \vdots & * \\ 1 & -3 & \vdots & * \\ -3 & 11 & \vdots & * \end{bmatrix}$$

e.g.  $\tilde{Q}_c = \begin{bmatrix} -2 & 7 & \vdots & \emptyset \\ 1 & -3 & \vdots & \emptyset \\ -3 & 11 & \vdots & -1 \end{bmatrix}$

$$\Rightarrow \det(\tilde{Q}_c) = 1$$

$$\Rightarrow \text{inv}(\tilde{Q}_c) = \text{adj}(\tilde{Q}_c) = \underbrace{\begin{bmatrix} 3 & 7 & \emptyset \\ 1 & 2 & \emptyset \\ 2 & 1 & -1 \end{bmatrix}}$$

$$\Rightarrow \hat{A} = T A T^{-1} = \begin{bmatrix} \emptyset & -2 & 22 \\ 1 & -3 & 2 \\ \emptyset & \emptyset & 4 \end{bmatrix}^T$$

$$\hat{\underline{b}} = T \cdot \underline{b} = \begin{bmatrix} \emptyset \\ \emptyset \\ \emptyset \end{bmatrix}$$

$$\hat{\underline{c}} = \underline{c}' T^{-1} = [1 \ -2 \ -1]$$

$$\Rightarrow \dot{\underline{z}} = \begin{bmatrix} \phi & -2 & 2 \\ 1 & -3 & 2 \\ \phi & \phi & 4 \end{bmatrix} \underline{z} + \begin{bmatrix} 1 \\ \phi \\ \phi \end{bmatrix} u$$

$$y = [1 \quad -2 \quad -4] \underline{z}$$

$$\Rightarrow \underline{\dot{z}}_3 = 4 z_3$$

Can obviously not be reached from the input. There exists an uncontrollable mode at

$$\underline{\lambda = 4.}$$

This mode is even unstable

$\Rightarrow$  The smallest disturbance will send some internal variables off into saturation.

$\Rightarrow$  Bad system.



In general: This transformation generates:

$$\dot{\underline{y}} = \begin{bmatrix} A_{11} & \vdots & A_{12} \\ \dots & \vdots & \dots \\ \emptyset & \vdots & A_{22} \end{bmatrix} \underline{y} + \begin{bmatrix} \underline{b}_1 \\ \vdots \\ \emptyset \end{bmatrix} u$$

↳ all these modes are uncontrollable.  $A_{22}$  contains the uncontrollable eigenvalues.

The controllable subsystem is:

$$\left| \begin{array}{l} \dot{\underline{x}}_c = A_{11} \cdot \underline{x}_c + \underline{b}_1 \cdot u \\ y = \underline{c}'_1 \cdot \underline{x}_c + d \cdot u \end{array} \right|$$

Example continued:

$$\left| \begin{array}{l} \dot{\underline{x}}_c = \begin{bmatrix} \emptyset & -2 \\ 1 & -3 \end{bmatrix} \underline{x}_c + \begin{bmatrix} 1 \\ \emptyset \end{bmatrix} u \\ y = [1 \quad -2] \underline{x}_c \end{array} \right|$$

We must now also get rid of those modes which are unobservable.

Algorithm:

- (1) Build the observability matrix:

$$Q_0 = [C'; C'A; \dots; C'A^{n-1}]$$

- (2) If  $\det(Q_0) \equiv 0 \Rightarrow$  there exist unobservable modes.

- (3) Take as many rows as are linearly independent, and extend from below with something that gives a full rank

$$\tilde{Q}_0 = \begin{bmatrix} Q_0 \\ \text{lin. indep.} \\ * \end{bmatrix}$$

(4) Use  $Q_0^{-1}$  as  $T$  for a similarity transformation.

Example continued:

$$Q_0 = \begin{bmatrix} 1 & -2 \\ -2 & 4 \end{bmatrix} \Rightarrow \det(Q_0) = 0$$

$$\Rightarrow \tilde{Q}_0 = \begin{bmatrix} 1 & -2 \\ \vdots & \vdots \\ 0 & 1 \end{bmatrix} \Rightarrow \det(\tilde{Q}_0) = 1$$

$$\Rightarrow \tilde{A} = T \cdot A_c / T = \begin{bmatrix} -2 & 0 \\ 1 & -1 \end{bmatrix}$$

$$\tilde{b} = T \cdot b_c = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$\tilde{c}' = c'_c / T = [1 \ 0]$$

$$\Rightarrow \tilde{y} = \begin{bmatrix} -2 & \vdots & 0 \\ \vdots & \vdots & \vdots \\ 1 & \vdots & -1 \end{bmatrix} \tilde{y} + \begin{bmatrix} 1 \\ \vdots \\ 0 \end{bmatrix} u$$

$$y = [1 \ \vdots \ 0] \tilde{y}$$

The second state-variable does not directly go into the output neither does it influence any of the other states

$\Rightarrow \gamma_2$  is unobservable.

In general: This transformation gives a representation:

$$\left| \begin{array}{l} \dot{\underline{\gamma}} = \begin{bmatrix} A_{11} & \vdots & \Phi \\ \vdots & \ddots & \vdots \\ A_{21} & \vdots & A_{22} \end{bmatrix} \underline{\gamma} + \begin{bmatrix} \underline{b}_1 \\ \vdots \\ \underline{b}_2 \end{bmatrix} u \\ \underline{y} = \begin{bmatrix} \underline{c}_1 & \vdots & \Phi \end{bmatrix} \underline{\gamma} + \underline{d} u \end{array} \right|$$

all these states are unobservable,

$A_{22}$  contains the unobservable eigenvalues.

$$\Rightarrow \left| \begin{array}{l} \dot{\underline{\gamma}}_1 = A_{11} \cdot \underline{\gamma}_1 + \underline{b}_1 \cdot u \\ \underline{y} = \underline{c}_1 \cdot \underline{\gamma}_1 + d \cdot u \end{array} \right|$$

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is the observable subsystem.

Example continued:

There is a stable unobservable eigenvalue at

$$\underline{\underline{\lambda = -1}}$$

The remaining subsystem is:

$$\left| \begin{array}{l} \dot{x}_1 = -2x_1 + u \\ y = x_1 \end{array} \right|$$

$$\Rightarrow \underline{\underline{G(s) = \frac{1}{s+2}}}$$

From the last example, a number of observations can be made:

- (1) While we have shown earlier that similar state-space representations always show the same transfer function, it is incorrect to conclude that all possible representations of a system with the same transfer function are similar to each other. (The third order representation is not related by a similarity transformation to the first order representation.)
- (2) The presence of uncontrollable or unobservable modes leads always to pole/zero-cancellation in the transfer function. It is correct (at least for SISO systems!) that pole/zero-cancellations in the transfer function always indicate the presence of either uncontrollable

- (3) Among the state-space representations there exists one important subclass namely containing all those representations in which all uncontrollable and unobservable modes have been eliminated. These are the representations of minimal degree and they are often referred to as minimal realizations.
- (4) All minimal realizations of a system are related by similarity transformations.
- (5) Iff the  $Q_c$ -matrix of a system is non-singular ( $\det(Q_c) \neq 0$ ), there do not exist any uncontrolled modes.
- (6) Iff the  $Q_o$ -matrix of a system is non-singular, there are no unobservable modes.
- (7) A controller-canonical realization of a system exists if the system is fully controllable.

(8) The number of linearly independent columns of the  $Q_c$ -matrix determine the number of controllable modes. This is also called the Rank of the  $Q_c$ -matrix. The Rank of a matrix can e.g. be determined by the following algorithm:

Try to invert  $Q_c$  by a Gaussian elimination. The number of pivot steps that can be performed before the remaining submatrix is all zero is equal to the rank of the matrix.

We shall often write:

- $\text{ord}(Q_c) \quad \therefore$  dimension of a square matrix  $Q_c$
- $\rho_c(Q_c) \quad \therefore$  Rank of the  $Q_c$ -matrix
- $\nu_c(Q_c) \quad \therefore$  Rank deficiency or Nullicity of the  $Q_c$ -matrix.

Obviously:  $\rho_c(Q_c) + \nu_c(Q_c) \equiv \text{ord}(Q_c)$

$\uparrow$   
column



(9) If in a representation there exist a number of rows of  $[A, b]$  which are zero except of the diagonal block of  $A \Rightarrow$  all these states are uncontrollable

$$\left| \begin{array}{l} \dot{x} = \begin{bmatrix} A_1 & * & A_2 \\ \phi & \boxed{*} & \phi \\ A_3 & * & A_4 \end{bmatrix} x + \begin{bmatrix} b_1 \\ \phi \\ b_2 \end{bmatrix} u \\ y = [c_1 \quad \dots \quad * \quad \dots \quad c_2] x + [d] u \end{array} \right|$$

More Controllable subsystem (if no other states are uncontrollable):

$$\left| \begin{array}{l} \dot{\underline{x}} = \begin{bmatrix} A_1 & A_2 \\ A_3 & A_4 \end{bmatrix} \underline{x} + \begin{bmatrix} b_1 \\ \vdots \\ b_2 \end{bmatrix} u \\ y = [c_1 \quad \dots \quad c_2] \underline{x} + [d] u \end{array} \right|$$

This can of course be a little generalized:

$$\left| \begin{array}{l} \dot{x} = \begin{bmatrix} -1 & \phi & 25 & -18 & -13 & 2 \\ \phi & \textcircled{-2} & \phi & \phi & \textcircled{1} & \phi \\ 23 & 14 & 14 & -1 & \phi & -1 \\ 2 & 22 & 18 & -10 & \phi & -5 \\ \phi & \textcircled{2} & \phi & \phi & \textcircled{8} & \phi \\ 18 & \phi & 23 & 27 & 5 & -1 \end{bmatrix} x + \begin{bmatrix} \phi \\ \phi \\ -1 \\ \phi \\ -1 \end{bmatrix} u \\ y = [1 \quad 2 \quad 3 \quad \phi \quad 4 \quad \phi] x + [27] u \end{array} \right| \begin{array}{l} \leftarrow \\ \leftarrow \end{array}$$

$$\left| \begin{array}{l} \dot{x}_2 = -2x_2 + x_5 \\ \dot{x}_5 = 2x_2 - 8x_5 \end{array} \right| \Rightarrow \text{uncontrollable}$$

$\begin{bmatrix} -2 & 1 \\ 2 & -8 \end{bmatrix}$  is such a "diagonal block"  
 $\Rightarrow$  States 2 and 5 are uncontrollable  
 We can cancel the 2nd and 5th  
 row and column

$$\Rightarrow \left| \begin{array}{l} \dot{\underline{x}}_c = \begin{bmatrix} -1 & 25 & -18 & 2 \\ 23 & 14 & -1 & -1 \\ 2 & 28 & -10 & -5 \\ 18 & 23 & 27 & -1 \end{bmatrix} \underline{x}_c + \begin{bmatrix} \phi \\ 1 \\ -1 \\ 1 \end{bmatrix} u \\ y = [1 \quad 3 \quad \phi \quad \phi] \underline{x}_c + [27] u \end{array} \right|$$

is the controllable subsystem (unless there exist further uncontrollable modes that are yet invisible as such).

(10) One convenient way to get rid of all uncontrollable modes is the input-decoupling algorithm shown before which transforms the system into:

$$\left| \begin{array}{l} \dot{\underline{z}} = \begin{bmatrix} A_c & \vdots & A_r \\ \hline \phi & \vdots & A_c \end{bmatrix} \underline{z} + \begin{bmatrix} b_c \\ \vdots \\ \phi \end{bmatrix} u \\ y = [c'_1 \quad \vdots \quad c'_2] \underline{z} + [d] u \end{array} \right|$$

$A_c$  contains the controllable modes  
 $A_{\bar{c}}$  contains the uncontrollable modes

$$\Rightarrow \left| \begin{array}{l} \dot{\underline{x}}_c = [A_c] \underline{x}_c + [b_c] u \\ y = [c_c] \underline{x}_c + [d] u \end{array} \right|$$

is the controllable subsystem which has no uncontrollable modes left.

(ii) The number of linearly independent rows of the  $Q_0$ -matrix determines the number of observable modes, thus:

$$\begin{aligned} \mathcal{S}_r(Q_0) &:: \# \text{ observable modes} \\ \mathcal{V}_r(Q_0) &:: \# \text{ unobservable modes.} \\ &\quad \leftarrow \text{row} \end{aligned}$$

[ However, in a square matrix :  
 $\mathcal{S}_r(\cdot) \equiv \mathcal{S}_c(\cdot)$  . ]

(12) If in a representation there exist a number of columns of  $[A; \underline{c}']$  which are zero except for the "diagonal" block of  $A \Rightarrow$  all these states are unobservable

$$\left| \begin{array}{l} \dot{\underline{x}} = \begin{bmatrix} A_1 & | & \phi & | & A_2 \\ \hline * & | & * & | & * \\ \hline A_3 & | & \phi & | & A_4 \end{bmatrix} \underline{x} + \begin{bmatrix} \underline{b}_1 \\ \vdots \\ \underline{b}_2 \end{bmatrix} u \\ y = [ \underline{c}' \quad | \quad \phi \quad | \quad \underline{c}'' ] \underline{x} + [d] u \end{array} \right|$$

$\Rightarrow$  Reduced subsystem with these unobservable modes removed:

$$\left| \begin{array}{l} \dot{\underline{z}} = \begin{bmatrix} A_1 & | & A_2 \\ \hline A_3 & | & A_4 \end{bmatrix} \underline{z} + \begin{bmatrix} \underline{b}_1 \\ \vdots \\ \underline{b}_2 \end{bmatrix} u \\ y = [ \underline{c}' \quad | \quad \underline{c}'' ] \underline{z} + [d] u \end{array} \right|$$

(13) One convenient way to get rid of all unobservable modes is the output-decoupling algorithm shown before which transforms the system into:

$$\left. \begin{aligned} \dot{\underline{x}} &= \begin{bmatrix} A_0 & | & \phi \\ \hline A_{21} & | & A_{20} \end{bmatrix} \underline{x} + \begin{bmatrix} b_0 \\ \vdots \\ b_2 \end{bmatrix} u \\ y &= [c_0' \quad \phi] \underline{x} + [d] u \end{aligned} \right| \quad -49-$$

$A_0$  contains the observable modes, and  
 $A_{20}$  contains the unobservable modes

$$\Rightarrow \dot{x}_0 = [A_0] x_0 + [b_0] u$$

$$y = [c_0'] x_0 + [d_r] u$$

is the observable subsystem which  
 has no unobservable modes left.

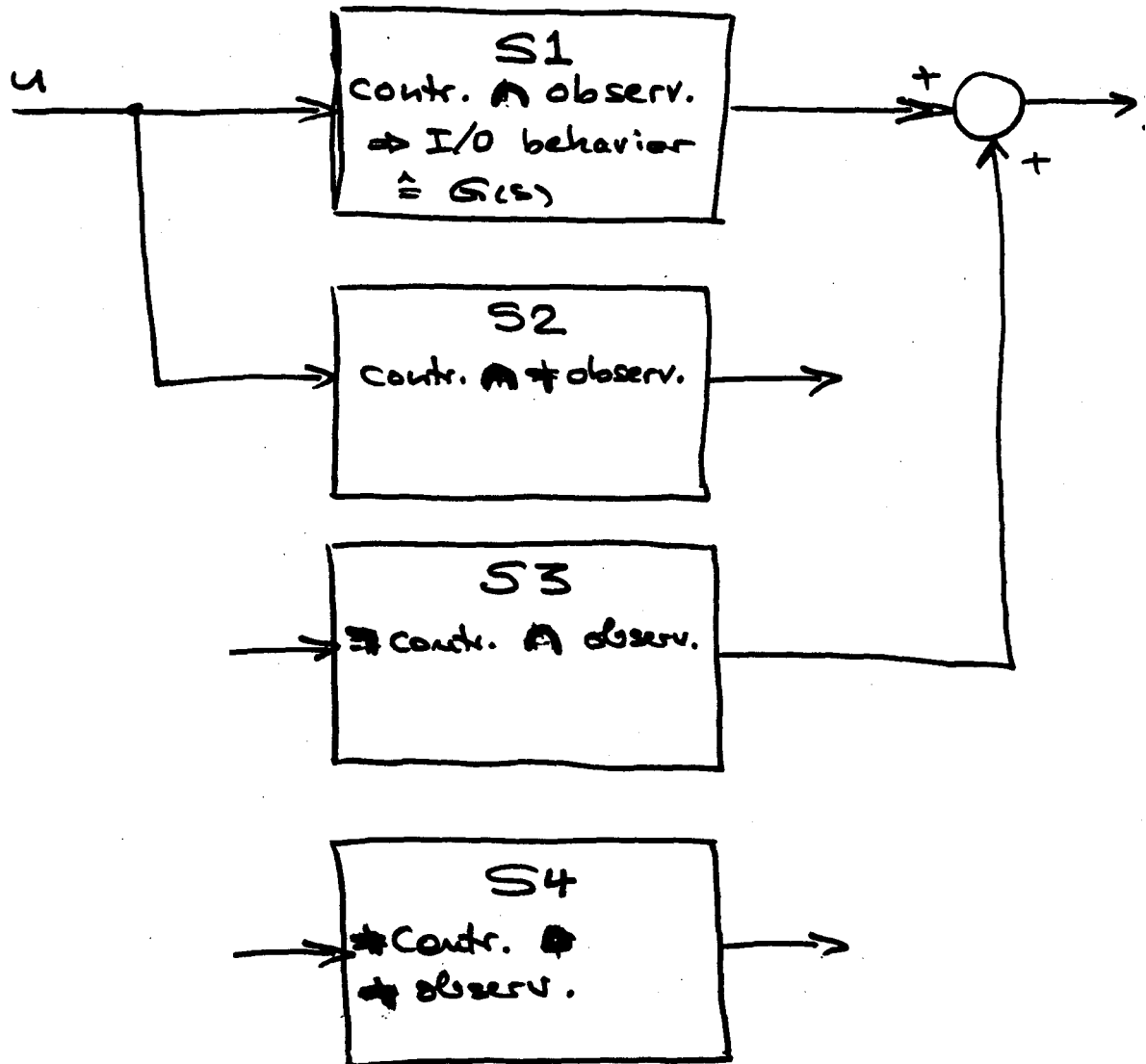
(14) Uncontrollable/unobservable modes are not visible from the outside. They do not influence the input/output behavior (that is why they don't show in the transfer function). They influence internal state behavior only.

(15) Uncontrollable/unobservable modes are normally harmless (just wasteful). However, this is true only if they are all stable:

- If there exist instable uncontrollable / unobservable modes, the smallest disturbance will drive some internal state variables into saturation. Saturation is a non-linear element. The system will at this moment cease to behave like a linear system, and the formerly uncontrollable / unobservable modes become suddenly visible and will make the performance of the system unusable.
- A system with instable uncontrollable unobservable modes is to be trashed. There is nothing that we can do to make this system work correctly.

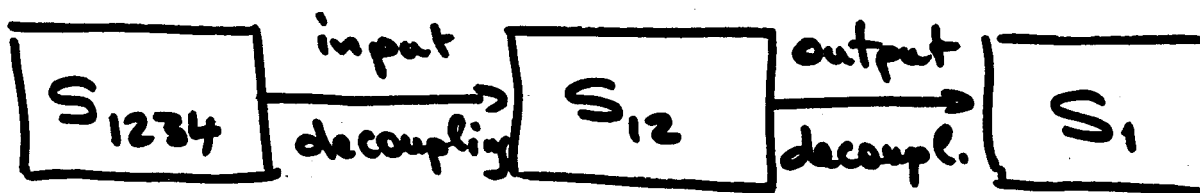
# The Kalman Decomposition

Any system can be decomposed in the following manner:



whereby any and all of these subsystems may be present / missing.

- The transfer function contains the subsystem  $S_1$  only, all others cancel out (pole/zero cancellation). The same is true for any minimal realization.
- Given a particular state-space representation, a minimal realization can be found by input/output decoupling:



or:

