

Solution of Linear Systems

A) In the time domain

Given: $\begin{vmatrix} \dot{\underline{x}} = A\underline{x} + B\underline{u} \\ \underline{y} = C\underline{x} + D\underline{u} \end{vmatrix}; \begin{matrix} \underline{x}(t=\phi) = \underline{x}_0 \\ \underline{u}(t)=\emptyset, \forall t < \phi \end{matrix}$

$$\Rightarrow \underline{x}(t) = e^{At} \cdot \underline{x}_0 + \int_0^t e^{A(t-\tau)} \cdot B \cdot \underline{u}(\tau) d\tau$$

$$\Rightarrow \underline{y}(t) = C \cdot e^{At} \cdot \underline{x}_0 + C \int_0^t e^{A(t-\tau)} \cdot B \cdot \underline{u}(\tau) d\tau + D \cdot \underline{u}(t)$$

B) In the frequency domain

$$\dot{\underline{x}} \rightarrow s \cdot \underline{x}(s) - \underline{x}_0$$

$$\Rightarrow s \cdot \underline{x}(s) - \underline{x}_0 = A \cdot \underline{x}(s) + B \cdot \underline{u}(s)$$

$$\Rightarrow s \cdot \underline{x}(s) - A \cdot \underline{x}(s) = \underline{x}_0 + B \cdot \underline{u}(s)$$

$$\Rightarrow s \cdot I^{(n)} \cdot \underline{x}(s) - A \cdot \underline{x}(s) = \underline{x}_0 + B \cdot \underline{u}(s)$$

\uparrow Identity matrix of dimensions
 $n \times n$

$$\Rightarrow [s \cdot I^{(n)} - A] \cdot \underline{x}(s) = \underline{x}_0 + B \cdot \underline{u}(s)$$

- 33 -

$$\Rightarrow \underline{X}(s) = [sI^{(n)} - A]^{-1} \cdot \underline{x}_0 + [sI^{(n)} - A]^{-1} \cdot B \cdot \underline{u}(s)$$

$$\Rightarrow \underline{y}(s) = C \cdot [sI^{(n)} - A]^{-1} \cdot \underline{x}_0 + C \cdot [sI^{(n)} - A]^{-1} \cdot B \cdot \underline{u}(s) + D \cdot \underline{u}(s)$$

c) Comparison of time- and frequency-domain solutions:

$$C \cdot e^{At} \cdot \underline{x}_0 \longrightarrow C \cdot [sI^{(n)} - A]^{-1} \cdot \underline{x}_0$$

$$\Rightarrow e^{At} \longrightarrow [sI^{(n)} - A]^{-1}$$

$$\Rightarrow e^{At} = \mathcal{F}^{-1} \left\{ (sI^{(n)} - A)^{-1} \right\}$$

Example:

$$A = \begin{bmatrix} -8 & 2 \\ -15 & 3 \end{bmatrix}$$

$$\Rightarrow sI^{(2)} - A = \begin{bmatrix} s & 0 \\ 0 & s \end{bmatrix} - \begin{bmatrix} -8 & 2 \\ -15 & 3 \end{bmatrix}$$

$$= \begin{bmatrix} (s+8) & -2 \\ 15 & (s-3) \end{bmatrix}$$

- 34 -

$$\begin{aligned} [sI^{(2)} - A]^{-1} &= \frac{[sI^{(2)} - A]^+}{|sI^{(2)} - A|} \\ &= \frac{\text{adj}(sI^{(2)} - A)}{\det(sI^{(2)} - A)} \end{aligned}$$

$$[sI^{(2)} - A]^+ = \begin{bmatrix} (s-3) & 2 \\ -15 & (s+8) \end{bmatrix}$$

$$\begin{aligned} |sI^{(2)} - A| &= (s+8)(s-3) - 15 \cdot (-2) \\ &= s^2 + 5s - 24 + 30 \\ &= s^2 + 5s + 6 \\ &= (s+2)(s+3) \end{aligned}$$

$$\Rightarrow [sI^{(2)} - A]^{-1} = \begin{bmatrix} \frac{s-3}{(s+2)(s+3)} & \frac{2}{(s+2)(s+3)} \\ \frac{-15}{(s+2)(s+3)} & \frac{s+8}{(s+2)(s+3)} \end{bmatrix}$$

We need to use partial fraction expansion on all four terms:

$$\frac{s-3}{(s+2)(s+3)} = \frac{a_{11}}{s+2} + \frac{b_{11}}{s+3}$$

$$a_{11} = \lim_{s \rightarrow -2} (s+2) \cdot \frac{s-3}{(s+2)(s+3)}$$

$$= \lim_{s \rightarrow -2} \frac{s-3}{s+3} = \frac{-5}{1} = -5$$

$$b_{11} = \lim_{s \rightarrow -3} (s+3) \cdot \frac{s-3}{(s+2)(s+3)}$$

$$= \lim_{s \rightarrow -3} \frac{s-3}{s+2} = \frac{-6}{-1} = 6$$

Similarly:

$$\frac{2}{(s+2)(s+3)} = \frac{2}{s+2} + \frac{-2}{s+3}$$

$$\frac{-15}{(s+2)(s+3)} = \frac{-15}{s+2} + \frac{15}{s+3}$$

$$\frac{s+8}{(s+2)(s+3)} = \frac{6}{s+2} + \frac{-5}{s+3}$$

$$\Rightarrow [S^{-1} - A]^{-1} = \begin{bmatrix} \left(\frac{-5}{s+2} + \frac{6}{s+3}\right) & \left(\frac{2}{s+2} + \frac{-2}{s+3}\right) \\ \left(\frac{-15}{s+2} + \frac{15}{s+3}\right) & \left(\frac{6}{s+2} + \frac{-5}{s+3}\right) \end{bmatrix}$$

$$\Rightarrow e^{At} = \mathcal{L}^{-1} \left\{ [sI - A]^{-1} \right\}$$
$$= \begin{bmatrix} (-5e^{-2t} + 6e^{-3t}) & (2e^{-2t} - 2e^{-3t}) \\ (-15e^{-2t} + 15e^{-3t}) & (6e^{-2t} - 5e^{-3t}) \end{bmatrix}$$
$$\neq \begin{bmatrix} e^{-8t} & e^{2t} \\ e^{-15t} & e^{3t} \end{bmatrix}$$

In Matlab:

$$A = [-8, 2; -15, 3]$$

$$e^A = \text{expm}(A)$$

↑ matrix exponential

SISO Systems & Transfer Functions:

Given a single-input / single-output (SISO) system:

$$\left| \begin{array}{l} \dot{x} = A \cdot x + b u \\ y = C \cdot x + d u \end{array} \right|; \quad \begin{array}{l} x(t=0) = x_0 \\ u(t) = \phi, \forall t < \infty \end{array}$$

$$y(t) = \underline{c}' \cdot e^{\underline{A}t} \cdot \underline{x}_0 + \underline{c}' \cdot \int_{0^-}^{A(t-\tau)} \underline{b} \cdot u(\tau) d\tau + d \cdot u(t)$$

$$Y(s) = C' [sI^{(n)} - A]^{-1} \cdot x_0 + C' [sI^{(n)} - A]^{-1} b \cdot U(s) + d \cdot U(s)$$

$$Y(s) = G(s) \cdot U(s) + \Gamma(s) \cdot x_0$$

↑ ↑
 transfer state transition
 function function

$$\Rightarrow \boxed{G(s) = C' (sI^n - A)^{-1} \cdot b + d}$$

$$\Gamma(s) = C'(sI^n - A)^{-1}$$

The transfer function $G(s)$ only accounts for the input, not for the initial condition.

Let: $x_0 = \emptyset$

$$\Rightarrow Y(s) = G(s) \cdot U(s)$$

Let: $u(t) = \delta(t)$

$$\Rightarrow U(s) = 1$$

$$\Rightarrow Y(s) = G(s) \cdot 1 = G(s)$$

$$y(t) = \mathcal{C} \int_0^t e^{At-t} \cdot b \cdot \delta(\tau) d\tau + d \cdot \delta(t)$$

$$= C e^{At} \cdot b + d \cdot \delta(t)$$

because of the sifting property of the Dirac distribution.

$$y(t) = \mathcal{F}^{-1} \{ G(s) \} = g(t) \quad \text{is the } \underline{\text{impulse response}}$$

$$\underline{g(t) = c'e^{At} \cdot b + d \cdot \delta(t)}$$

Let $G(s)$ be strictly_proper:

$$G(s) = \frac{N(s)}{D(s)} ; \text{ord}(N(s)) < \text{ord}(D(s))$$

$$\Leftrightarrow d = \emptyset$$

$$\Rightarrow g(t) = \mathcal{L}^{-1}\{G(s)\} = c'e^{At} \cdot b$$

We can write:

$$g(t-\tau) = c'e^{A(t-\tau)} \cdot b$$

and therefore:

$$y(t) = c' \cdot \int_{0^-}^t e^{A(t-\tau)} b \cdot u(\tau) d\tau$$

$$= \int_{0^-}^t c' \cdot e^{A(t-\tau)} b \cdot u(\tau) d\tau$$

$$= \int_{0^-}^t g(t-\tau) \cdot u(\tau) d\tau$$

- 4 φ -

To summarize:

Given a strictly proper SISO system with zero initial conditions:

$$\begin{cases} \dot{x} = Ax + bu \\ y = Cx \end{cases}; \quad \begin{aligned} x(t=0) &= \phi \\ u(t) &= \phi, \forall t < 0 \end{aligned}$$

$$\Rightarrow y(t) = C \int_{0^-}^t e^{A(t-\tau)} \cdot b \cdot u(\tau) d\tau \\ \equiv \int_{0^-}^t g(t-\tau) \cdot u(\tau) d\tau$$

where: $g(t) = f^{-1}\{G(s)\}$

and: $G(s) = \frac{Y(s)}{U(s)}$

and: $U(s) = f\{u(t)\}$
 $Y(s) = f\{y(t)\}$

- 41 -

We define:

$$Y(s) = G(s) \cdot U(s)$$

$$y(t) = g(t) * u(t)$$

↑ convolution operator

$$g(t) * u(t) := \int_{0^-}^t g(t-\tau) \cdot u(\tau) d\tau$$

↑ is defined as

Variable transformation:

$$\sigma = t - \tau$$

$$\Rightarrow \tau = t - \sigma$$

$$d\tau = -d\sigma$$

$$\begin{array}{c|c} \tau & \sigma \\ \hline 0^- & t \\ t & 0^- \end{array}$$

$$\Rightarrow g(t) * u(t) = \int_{0^-}^t g(t-\tau) \cdot u(\tau) d\tau$$

$$\begin{aligned} &= \int_t^{\infty} g(\sigma) \cdot u(t-\sigma) (-d\sigma) = \int_0^{\infty} u(t-\sigma) \cdot g(\sigma) d\sigma \\ &= u(t) * g(t) \end{aligned}$$

- 42 -

$$\Rightarrow \underline{g(t) * u(t) \equiv u(t) * g(t)}.$$

The (scalar) convolution operator
is commutative.

The convolution also works for
systems that are not strictly
proper, because:

$$\begin{aligned} y(t) &= g(t) * u(t) = \int_{-\infty}^t g(t-\tau) u(\tau) d\tau \\ &= \int_{-\infty}^t [e^{A(t-\tau)} b + d \cdot \delta(t-\tau)] u(\tau) d\tau \\ &= \int_{-\infty}^t e^{A(t-\tau)} b u(\tau) d\tau + \int_{-\infty}^t d \delta(t-\tau) u(\tau) d\tau \\ &= e^{\int_{-\infty}^t A(t-\tau)} b u(t) + d \cdot u(t) \end{aligned}$$

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