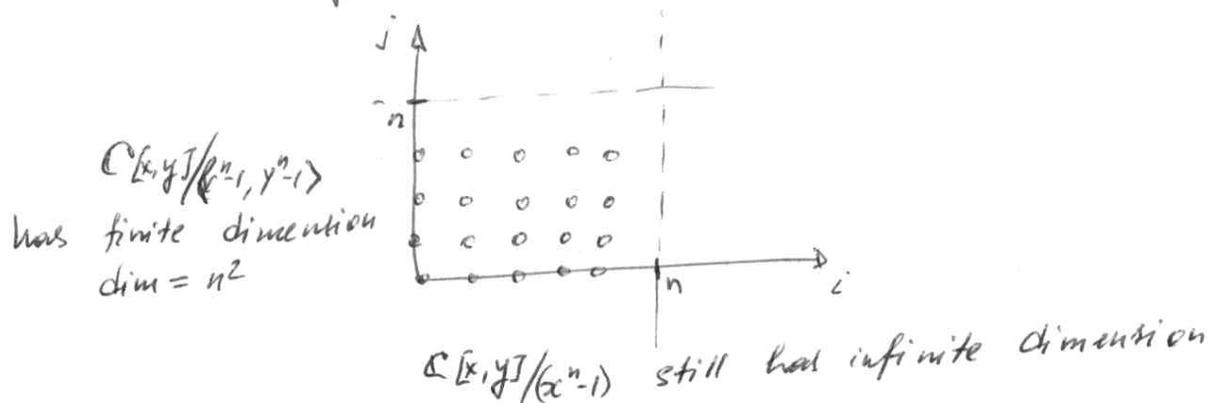


Finite shift-invariant 2D signal models

2D \Rightarrow 2 shifts, e.g. x and y .

Then $\mathcal{H} = \mathbb{C}[x, y] / \langle ? \rangle = \mathbb{C}[x, y] / \mathcal{I}$

- \mathcal{I} is an ideal: $\mathcal{I} = \langle p_1(x, y), \dots, p_k(x, y) \rangle$; $k \geq 2$ for finite-dimensional \mathcal{H} .
- $\mathbb{C}[x, y]$ is not a P.I.D., e.g. $\mathcal{I} = \langle x^2, y^2 \rangle$.
- \mathcal{I} can have arbitrary many generators; e.g. $\mathcal{I} = \langle x^{k-1}y, x^{k-2}y^2, \dots, xy^{k-1} \rangle$.
- \mathcal{I} needs ≥ 2 generators to have a finite basis; e.g. $\{x^i y^j\}$.



Assumption: We assume $\mathcal{I} = \langle p(x, y), q(x, y) \rangle$, where

$$p(x, y) = x^n + \sum_{l, k=0}^{n-1} p_{l, k} x^l y^k$$

$$q(x, y) = y^n + \sum_{l, m=0}^{n-1} q_{m, l} x^l y^m$$

Then $\dim \mathcal{H} = n^2$

Signal model

Assume finite 2D regular model:

$$A = M = \mathbb{C}[x, y] / \langle p(x, y), q(x, y) \rangle, \text{ basis } b = \{p_0(x, y), \dots, p_{n^2-1}(x, y)\}$$

$$\Phi: \mathbb{C}^{n^2} \rightarrow M \quad (\mathbb{C}^{n^2} \text{ is a space of } n^2 \times 1 \text{ vectors})$$
$$\hat{s} \mapsto \sum_{k=0}^{n^2-1} s_k p_k(x, y)$$

Example (for the whole lecture):

$$A = M = \mathbb{C}[x, y] / \langle x^{n-1}, y^{n-1} \rangle; \quad b = \{x^i y^j\}_{0 \leq i, j \leq n-1}$$

In general, $\Phi \hat{s} \mapsto \sum_{k=0}^{n^2-1} s_k p_k(x, y) = \sum_{k=0}^{n-1} \sum_{l=0}^{n-1} s_{nk+l} x^k y^l$

But in this case we consider \hat{s} an $n \times n$ array $[s_{k,l}]_{0 \leq k, l \leq n-1}$ (see example above); and space of signals is $\mathbb{C}^{n \times n}$ instead of \mathbb{C}^{n^2} .

Then $\Phi: \hat{s} \mapsto \sum_{k=0}^{n-1} \sum_{l=0}^{n-1} s_{k,l} x^k y^l$

Filtering

Multiplication $h(x, y) \cdot s(x, y) \text{ mod } \langle p(x, y), q(x, y) \rangle$

! This is well-defined only if I has a Groebner basis. However, the above assumption ensures that $\langle p(x, y), q(x, y) \rangle$ is a Groebner basis.

Example: $h(x, y) \cdot s(x, y) \text{ mod } \langle x^{n-1}, y^{n-1} \rangle$

This is a 2D circular convolution.

Why?

Filtering in coordinates

\mathcal{M} -basis b affords a representation φ of \mathcal{A} :

$$\varphi : \mathcal{A} \rightarrow \mathbb{C}^{n^2 \times n^2}$$

$$h \mapsto \varphi(h)$$

Then $h.s \text{ mod } \langle \dots \rangle \Leftrightarrow \varphi(h) \cdot \hat{s}$

Example : $b = \{xy^0, xy^1, \dots, xy^i, xy^{i+1}, \dots, x^{n-1}y^{n-1}\} = \{P_{intj}(x,y)\}$, $P_{intj} = x^i y^j$

$$\varphi(x \cdot P_{intj}) = x \cdot x^i y^j = x^{i+1} y^j = P_{(i+1)nj} = P_{intj} \circ \tau_n \Rightarrow$$

$$\Rightarrow \varphi(x) = \begin{pmatrix} \vdots & \vdots & \vdots & \vdots & \vdots \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ \vdots & \vdots & \vdots & \vdots & \vdots \end{pmatrix} = C_n \otimes I_n, \quad C_n = \begin{pmatrix} 0 & & & & 1 \\ & \ddots & & & \\ & & \ddots & & \\ & & & \ddots & \\ & & & & 0 \end{pmatrix} \text{ - circular shift}$$

$$y \cdot P_{intj} = y \cdot x^i y^j = x^i y^{j+1} = P_{int(j+1)} \Rightarrow$$

$$\Rightarrow \varphi(y) = \begin{pmatrix} 0 & \vdots & \vdots & \vdots & \vdots \\ \vdots & 0 & \vdots & \vdots & \vdots \\ \vdots & \vdots & 0 & \vdots & \vdots \\ \vdots & \vdots & \vdots & 0 & \vdots \\ \vdots & \vdots & \vdots & \vdots & 0 \end{pmatrix} = I_n \otimes C_n$$

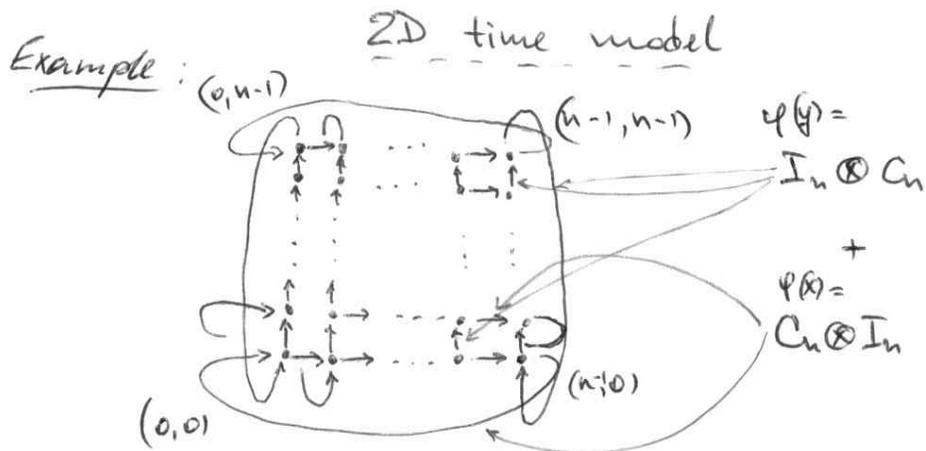
$$\text{Then } \varphi(h) = h(\varphi(x), \varphi(y)) = \sum h_{k,e} (C_n \otimes I_n)^k (I_n \otimes C_n)^e = \sum h_{k,e} (C_n^k \otimes I_n) (I_n \otimes C_n^e) =$$

$$= \sum h_{k,e} C_n^k \otimes C_n^e.$$

(Recall that $(A \otimes B)(C \otimes D) = AC \otimes BD$).

Visualization

We define visualization of a signal model as a graph with adjacency matrix $\varphi(x) + \varphi(y)$.



This is a torus
(a 'doughnut')

Spectrum and Fourier Transform

Assume that $p(x, y) = g(x, y) = 0$ has n^2 solutions $\alpha = \{(d_k, \beta_k)\}_{0 \leq k \leq n^2-1}$

CRT:

$$\Delta : M \rightarrow \bigoplus_{k=0}^{n^2-1} \mathbb{C}[x, y] / \langle x - d_k, y - \beta_k \rangle$$

$$s(x, y) \mapsto (s(d_0, \beta_0), \dots, s(d_{n^2-1}, \beta_{n^2-1}))$$

Pure frequency: $f_k = \Delta^{-1}(e_k), 0 \leq k \leq n^2-1$

In coordinates:

Basis in M : b (fixed by the signal model)

Basis in each spectral component: $\{1\}$

$$F = P_{b, \alpha} = [P_e(d_k, \beta_k)]_{0 \leq k, e \leq n^2-1} \in \mathbb{C}^{n^2 \times n^2}$$

if bases other than $\{1\} \Rightarrow F' = D \cdot F$

Example: $x^{n-1} = y^{n-1} = 0$ has n^2 solutions $\alpha = \{ (\omega_n^k, \omega_n^l) \}_{0 \leq k, l \leq n-1}$

$$\alpha_{kn+1} = (\omega_n^k, \omega_n^l)$$

Basis $b = \{ P_{\alpha} \}_{0 \leq l \leq n-1}, P_{\alpha} = x^i y^j$

$$\begin{aligned} \text{Thus, } F = P_{b, \alpha} &= [P_{\alpha}(d_m)]_{0 \leq t, m \leq n-1} = [P_{\alpha}(j, k)]_{0 \leq i, j, k, l \leq n-1} \\ &= [\omega_n^{ik+lj}]_{0 \leq i, j, k, l \leq n-1} \end{aligned}$$

(i, l running first)

$$= \begin{pmatrix} 1 & 1 & \dots & 1 \\ \omega_n & \omega_n^2 & \dots & \omega_n^{n-1} \\ \omega_n^2 & \omega_n^4 & \dots & \omega_n^{2(n-1)} \\ \vdots & \vdots & \ddots & \vdots \\ \omega_n^{(n-1)} & \omega_n^{2(n-1)} & \dots & \omega_n^{2(n-1)(n-1)} \end{pmatrix} = \begin{pmatrix} \omega_n^0 \cdot \text{DFT}_n & \omega_n^0 \cdot \text{DFT}_n & \dots & \omega_n^0 \cdot \text{DFT}_n \\ \omega_n^1 \cdot \text{DFT}_n & \omega_n^1 \cdot \text{DFT}_n & \dots & \omega_n^1 \cdot \text{DFT}_n \\ \vdots & \vdots & \ddots & \vdots \\ \omega_n^{n-1} \cdot \text{DFT}_n & \omega_n^{n-1} \cdot \text{DFT}_n & \dots & \omega_n^{(n-1)^2} \cdot \text{DFT}_n \end{pmatrix} =$$

$$= \text{DFT}_n \otimes \text{DFT}_n$$

Frequency response

Each spectral component $\mu_k = \mathbb{C}[x, y] / \langle x - \alpha_k, y - \beta_k \rangle$ with basis $\{1\}$ affords an irreducible representation

ψ_k of \mathcal{A} :

$$\psi_k : \mathcal{A} \rightarrow \mathbb{C}^{1 \times 1}$$

$$h(x, y) \mapsto h(\alpha_k, \beta_k) \quad (\equiv h(x, y) \cdot 1 \pmod{\langle x - \alpha_k, y - \beta_k \rangle})$$

$(h(\alpha_k, \beta_k))_{0 \leq k \leq n^2-1}$ is the frequency response of $h(x, y)$.

it implies

$$F \psi(h) F^{-1} = \text{diag} (h(\alpha_k, \beta_k))_{0 \leq k \leq n^2-1}$$

Commutative diagram:

$$\begin{array}{ccc}
 \mathbb{C}[x, y] / \langle p, q \rangle & \xrightarrow{F} & \bigoplus \mathbb{C}[x, y] / \langle x - \alpha_k, y - \beta_k \rangle \\
 \downarrow \varphi(h) & & \downarrow \text{diag}_k (h(\alpha_k, \beta_k)) \\
 \mathbb{C}[x, y] / \langle p, q \rangle & \xrightarrow{F} & \bigoplus \mathbb{C}[x, y] / \langle x - \alpha_k, y - \beta_k \rangle
 \end{array}$$

Example:

$$F = \text{DFT}_n \otimes \text{DFT}_n$$

$$F \text{ diagonalizes } \varphi(x) : F \varphi(x) F^{-1} = (\text{DFT}_n \otimes \text{DFT}_n) (C_n \otimes I_n) (\text{DFT}_n \otimes \text{DFT}_n)^{-1} =$$

$$= (\text{DFT}_n \otimes \text{DFT}_n) (C_n \otimes I_n) (\text{DFT}_n^{-1} \otimes \text{DFT}_n^{-1}) =$$

$$= \text{DFT}_n C_n \text{DFT}_n^{-1} \otimes \text{DFT}_n I_n \text{DFT}_n^{-1} = \text{diag}_{0 \leq k \leq n-1} (w_n^k) \otimes I_n$$

F diagonalizes $\varphi(y)$:

$$F \varphi(y) F^{-1} = (\text{DFT}_n \otimes \text{DFT}_n) (I_n \otimes C_n) (\text{DFT}_n^{-1} \otimes \text{DFT}_n^{-1}) =$$

$$= \text{DFT}_n I_n \text{DFT}_n \otimes \text{DFT}_n C_n \text{DFT}_n^{-1} =$$

$$= I_n \otimes \text{diag}_{0 \leq k \leq n-1} (w_n^k)$$

Hence, F diagonalizes $\varphi(x) + \varphi(y)$.

Note: F diagonalizes $\varphi(x) + \varphi(y)$ is a "weaker" statement than F diagonalizes $\varphi(x)$ and $\varphi(y)$.

Separable vs non-separable SP

In our example: $A = M = \mathbb{C}[x, y] / \langle p(x), p(y) \rangle$, $p(x) = x^n - 1$

This is a separable case $\Rightarrow F$ is a tensor product of corresponding ID F .

In general, $A = M = \mathbb{C}[x, y] / \langle p(x, y), q(x, y) \rangle$ - non-separable